

ISL28108, ISL28208, ISL28408

40V Precision Single Supply Rail-to-Rail Output Low Power Operational Amplifiers

The ISL28108, ISL28208 and ISL28408 are single, dual and quad low power precision amplifiers optimized for single supply applications. These devices feature a common mode input voltage range extending to 0.5V below the V- rail, a rail-to-rail differential input voltage range for use as a comparator, and rail-to-rail output voltage swing, which make them ideal for single supply applications where input operation at ground is important.

Added features include low offset voltage, and low temperature drift making them the ideal choice for applications requiring high DC accuracy. The output stage is capable of driving large capacitive loads from rail-to-rail for excellent ADC driving performance. The devices can operate for single or dual supply from 3V ($\pm 1.5V$) to 40V ($\pm 20V$) and are fully characterized at $\pm 5V$ and $\pm 15V$. The combination of precision, low power, and small footprint provide the user with outstanding value and flexibility relative to similar competitive parts.

Applications for these amplifiers include precision instrumentation, data acquisition, precision power supply control, and industrial control.

The ISL28108 single is offered in 8 Ld TDFN, MSOP and SOIC packages. The ISL28208 dual amplifier is offered in 8 Ld TDFN, MSOP, and SOIC packages. The ISL28408 is offered in 14 Ld SOIC package. All devices are offered in standard pin configurations and operate over the extended temperature range of -40°C to $+125^{\circ}\text{C}$.

Features

- Single or dual supply, rail-to-rail output and below ground (V-) input capability
- Rail-to-rail input differential voltage range for comparator applications
- Single supply range 3V to 40V
- Low current consumption ($V_S = \pm 5V$)..... 165 μA
- Low noise voltage..... 15.8nV/ $\sqrt{\text{Hz}}$
- Low noise current..... 80fA/ $\sqrt{\text{Hz}}$
- Low input offset voltage (ISL28108)..... 150 μV
- Superb temperature drift
 - Voltage offset TC 0.1 $\mu\text{V}/^{\circ}\text{C}$, Typ
- Low input bias current..... -13nA Typ
- Operating temperature range..... -40°C to $+125^{\circ}\text{C}$
- No phase reversal

Applications

- Precision instruments
- Medical instrumentation
- Data acquisition
- Power supply control
- Industrial process control

Related Literature

- [AN1658](#), "ISL28208SOICEVAL2Z Evaluation Board User Guide"

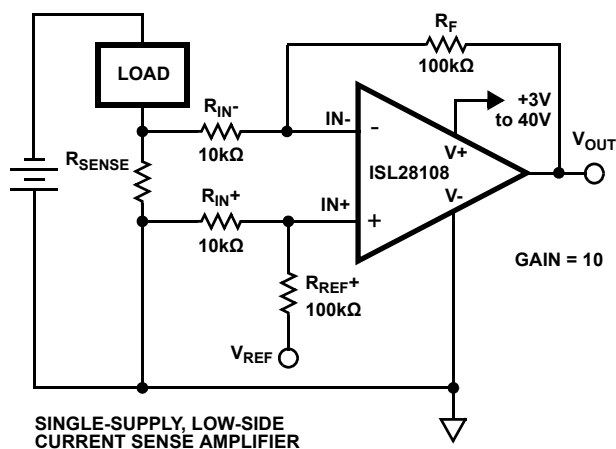


FIGURE 1. TYPICAL APPLICATION CIRCUIT

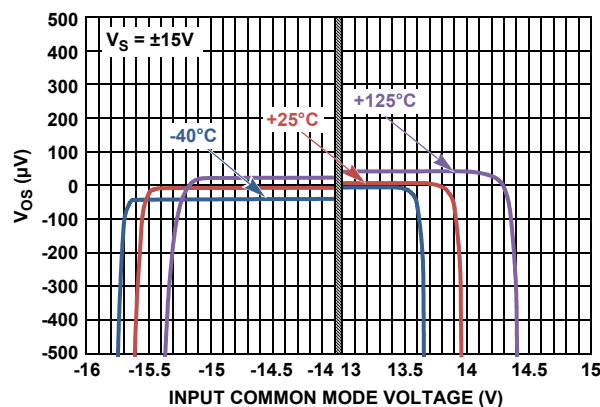


FIGURE 2. INPUT OFFSET VOLTAGE vs INPUT COMMON MODE VOLTAGE, $V_S = \pm 15V$

Table of Contents

Ordering Information	3
Pin Configurations	4
Pin Descriptions	5
Absolute Maximum Ratings	6
Thermal Information	6
Operating Conditions	6
Electrical Specifications, $V_S \pm 15$	6
Electrical Specifications, $V_S \pm 5V$	8
Typical Performance Curves	10
Applications Information	21
Functional Description	21
Operating Voltage Range	21
Input Stage Performance	21
Output Drive Capability	21
Output Phase Reversal	21
Unused Channels	22
Power Dissipation	22
ISL28108, ISL28208, ISL28408 SPICE Model	22
Characterization vs Simulation Results	25
Revision History	28
About Intersil	30
Package Outline Drawing, M8.15E	31
Package Outline Drawing, L8.3x3K	32
Package Outline Drawing, M8.118B	33
Package Outline Drawing, M14.15	34

Ordering Information

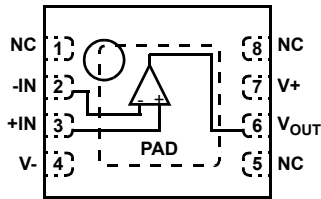
PART NUMBER (Notes 1, 2, 3)	PART MARKING	TEMP. RANGE (°C)	PACKAGE (Pb-Free)	PKG. DWG. #
ISL28108FBZ	28108 FBZ	-40 to +125	8 Ld SOIC	M8.15E
ISL28108FRTZ	108Z	-40 to +125	8 Ld TDFN	L8.3x3K
<i>Coming soon</i> ISL28108FUZ	8108Z	-40 to +125	8 Ld MSOP	M8.118B
ISL28208FBZ	28208 FBZ	-40 to +125	8 Ld SOIC	M8.15E
ISL28208FRTZ	208F	-40 to +125	8 Ld TDFN	L8.3x3K
ISL28208FUZ	8208Z	-40 to +125	8 Ld MSOP	M8.118B
ISL28408FBZ	28408 FBZ	-40 to +125	14 Ld SOIC	M14.15
ISL28208SOICEVAL2Z	Evaluation Board			

NOTES:

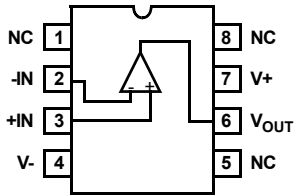
1. Add "-T*" suffix for tape and reel. Please refer to Tech Brief [TB347](#) for details on reel specifications.
2. These Intersil Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials, and 100% matte tin plate plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
3. For Moisture Sensitivity Level (MSL), please see device information page for [ISL28108](#), [ISL28208](#), [ISL28408](#). For more information on MSL please see Tech Brief [TB363](#).

Pin Configurations

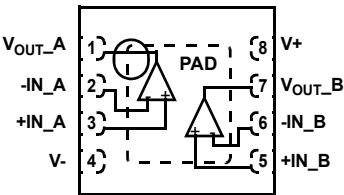
ISL28108
(8 LD TDFN)
TOP VIEW



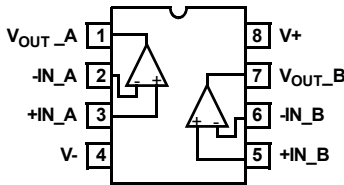
ISL28108
(8 LD MSOP, SOIC)
TOP VIEW



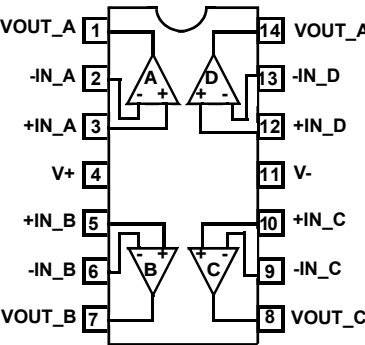
ISL28208
(8 LD TDFN)
TOP VIEW



ISL28208
(8 LD MSOP, SOIC)
TOP VIEW

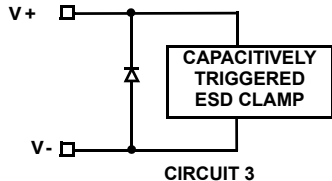
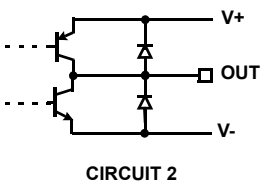
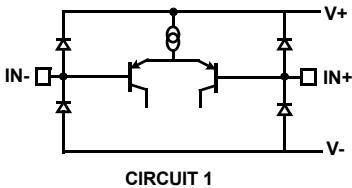


ISL28408
(14 LD SOIC)
TOP VIEW



Pin Descriptions

ISL28108 (8 Ld SOIC, MSOP, TDFN)	ISL28208 (8 Ld SOIC, MSOP, TDFN)	ISL28408 (14 Ld SOIC)	PIN NAME	EQUIVALENT CIRCUIT	DESCRIPTION
3	-	-	+IN	Circuit 1	Amplifier non-inverting input
-	3	3	+IN_A		
-	5	5	+IN_B		
-	-	10	+IN_C		
-	-	12	+IN_D		
4	4	11	V-	Circuit 3	Negative power supply
2	-	-	-IN	Circuit 1	Amplifier inverting input
-	2	2	-IN_A		
-	6	6	-IN_B		
-	-	9	-IN_C		
-	-	13	-IN_D		
7	8	4	V+	Circuit 3	Positive power supply
6	-	-	V _{OUT}	Circuit 2	Amplifier output
-	1	1	V _{OUT_A}		
-	7	7	V _{OUT_B}		
-	-	8	V _{OUT_C}		
-	-	14	V _{OUT_D}		
1, 5, 8	-	-	NC	-	No internal connection
PAD	PAD	-	PAD	-	Thermal Pad - TDFN package only. Connect thermal pad to ground or most negative potential.



Absolute Maximum Ratings

Maximum Supply Voltage	42V
Maximum Differential Input Voltage	42V or $V_- - 0.5V$ to $V_+ + 0.5V$
Min/Max Input Voltage	42V or $V_- - 0.5V$ to $V_+ + 0.5V$
Max/Min Input Current	$\pm 20mA$
Output Short-Circuit Duration (1 output at a time)	Indefinite
ESD Tolerance (ISL28208, ISL28408)	
Human Body Model (Tested per JESD22-A114F)	6kV
Machine Model (Tested per JESD22-A115-C)	400V
Charged Device Model (Tested per JESD22-C110D)	2kV
ESD Tolerance (ISL28108)	
Human Body Model (Tested per JESD22-A114F)	5.5kV
Machine Model (Tested per JESD22-A115-C)	300V
Charged Device Model (Tested per JESD22-C110D)	2kV
ESD Tolerance (ISL28108 MSOP package only)	
Human Body Model (Tested per JESD22-A114F)	3kV
Machine Model (Tested per JESD22-A115-C)	300V
Charged Device Model (Tested per JESD22-C110D)	2kV

Thermal Information

Thermal Resistance (Typical)	θ_{JA} ($^{\circ}C/W$)	θ_{JC} ($^{\circ}C/W$)
8 Ld SOIC Package (208, Notes 4, 7)	120	55
8 Ld SOIC Package (108, Notes 4, 7)	120	60
8 Ld TDFN Package (208, Notes 5, 6)	47	6
8 Ld TDFN Package (108, Notes 5, 6)	45	3.5
8 Ld MSOP Package (208, Notes 4, 7)	150	50
8 Ld MSOP Package (108, Notes 4, 7)	165	57
14 Ld SOIC Package (408, Notes 4, 7)	71	37
Storage Temperature Range	$-65^{\circ}C$ to $+150^{\circ}C$	
Pb-Free Reflow Profile	see link below	

Operating Conditions

Ambient Operating Temperature Range	$-40^{\circ}C$ to $+125^{\circ}C$
Maximum Operating Junction Temperature	$+150^{\circ}C$
Supply Voltage	3V ($\pm 1.5V$) to 40V ($\pm 20V$)

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

NOTES:

- θ_{JA} is measured with the component mounted on a high effective thermal conductivity test board in free air. See Tech Brief [TB379](#) for details.
- θ_{JA} is measured in free air with the component mounted on a high effective thermal conductivity test board with “direct attach” features. See Tech Brief [TB379](#).
- For θ_{JC} , the “case temp” location is the center of the exposed metal pad on the package underside.
- For θ_{JC} , the “case temp” location is taken at the package top center.

IMPORTANT NOTE: All parameters having Min/Max specifications are guaranteed. Typical values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_J = T_C = T_A$

Electrical Specifications, $V_S \pm 15$ $V_{CM} = 0$, $V_O = 0V$, $R_L = \text{Open}$, $T_A = +25^{\circ}C$, unless otherwise noted. **Boldface entries apply over the operating temperature range, $-40^{\circ}C$ to $+125^{\circ}C$. Temperature data established by characterization.**

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 8)	TYP	MAX (Note 8)	UNIT
V_{OS}	Input Offset Voltage	ISL28208	-230	25	230	μV
		ISL28408	-330		330	μV
		ISL28108 SOIC, TDFN	-150	10	150	μV
			-270		270	μV
TCV_{OS}	Input Offset Voltage Temperature Coefficient	ISL28208 SOIC $-40^{\circ}C$ to $+125^{\circ}C$		0.1	1.1	$\mu V/^{\circ}C$
		ISL28208 MSOP $-40^{\circ}C$ to $+125^{\circ}C$		0.2	1.5	$\mu V/^{\circ}C$
		ISL28208 TDFN ISL28408 $-40^{\circ}C$ to $+125^{\circ}C$		0.2	1.4	$\mu V/^{\circ}C$
		ISL28108 SOIC, TDFN $-40^{\circ}C$ to $+125^{\circ}C$		0.2	1.2	$\mu V/^{\circ}C$
ΔV_{OS}	Input Offset Voltage Match (ISL28208 only)	ISL28208 SOIC, TDFN	-300	5	300	μV
			-400		400	μV
		ISL28208 MSOP	-420		420	μV
I_B	Input Bias Current		-43	-13		nA
			-63			nA
TCI_B	Input Bias Current Temperature Coefficient			0.07		nA/ $^{\circ}C$

Electrical Specifications, $V_S \pm 15$ $V_{CM} = 0$, $V_O = 0V$, $R_L = \text{Open}$, $T_A = +25^\circ\text{C}$, unless otherwise noted. **Boldface entries apply over the operating temperature range, -40°C to $+125^\circ\text{C}$. Temperature data established by characterization. (Continued)**

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 8)	TYP	MAX (Note 8)	UNIT
I_{OS}	Input Offset Current	ISL28208	-3	0	3	nA
			-4		4	nA
		ISL28108 SOIC, TDFN ISL28408	-4	0	4	nA
			-5		5	nA
CMRR	Common-Mode Rejection Ratio	$V_{CM} = V_- - 0.5V$ to $V_+ - 1.8V$		119		dB
		$V_{CM} = V_- - 0.2V$ to $V_+ - 1.8V$		123		dB
				102		dB
		$V_{CM} = V_-$ to $V_+ - 1.8V$	105	123		dB
			102	115		dB
V_{CMIR}	Common Mode Input Voltage Range	Guaranteed by CMRR test	$V_- - 0.5$		$V_+ - 1.8$	V
			V_-		$V_+ - 1.8$	V
PSRR	Power Supply Rejection Ratio	$V_S = 3V$ to $40V$, $V_{CMIR} = \text{Valid Input Voltage}$	110	128		dB
			109	124		dB
A_{VOL}	Open-Loop Gain	$V_O = -13V$ to $+13V$, $R_L = 10k\Omega$ to ground	117	126		dB
			100			dB
V_{OL}	Output Voltage Low, V_{OUT} to V_-	$R_L = 10k\Omega$		52	85	mV
					145	mV
V_{OH}	Output Voltage High, V_+ to V_{OUT}	$R_L = 10k\Omega$		70	110	mV
					150	mV
I_S	Supply Current/Amplifier	$R_L = \text{Open}$		185	250	μA
				270	350	μA
I_{SC+}	Output Short Circuit Source Current	$R_L = 10\Omega$ to V_-		19		mA
I_{SC-}	Output Short Circuit Sink Current	$R_L = 10\Omega$ to V_+		30		mA
V_{SUPPLY}	Supply Voltage Range	Guaranteed by PSRR	3		40	V
AC SPECIFICATIONS						
GBWP	Gain Bandwidth Product	$A_{CL} = 101$, $V_O = 100mV_{P-P}$, $R_L = 2k\Omega$		1.2		MHz
e_{n-p-p}	Noise Voltage	0.1Hz to 10Hz; $V_S = \pm 18V$		580		nV-P-P
e_n	Noise Voltage Density	$f = 10\text{Hz}$; $V_S = \pm 18V$		18		nV/ $\sqrt{\text{Hz}}$
e_n	Noise Voltage Density	$f = 100\text{Hz}$; $V_S = \pm 18V$		16		nV/ $\sqrt{\text{Hz}}$
e_n	Noise Voltage Density	$f = 1\text{kHz}$; $V_S = \pm 18V$		15.8		nV/ $\sqrt{\text{Hz}}$
e_n	Noise Voltage Density	$f = 10\text{kHz}$; $V_S = \pm 18V$		15.8		nV/ $\sqrt{\text{Hz}}$
i_n	Noise Current Density	$f = 10\text{kHz}$; $V_S = \pm 18V$		80		fA/ $\sqrt{\text{Hz}}$
THD + N	Total Harmonic Distortion + Noise	1kHz, $A_V = 1$, $V_O = 3.5V_{RMS}$, $R_L = 10k\Omega$		0.00042		%
TRANSIENT RESPONSE						
SR	Slew Rate, V_{OUT} 20% to 80%	$A_V = 1$, $R_L = 2k\Omega$, $V_O = 10V_{P-P}$		0.45		V/ μs
t_r , t_f , Small Signal	Rise Time, V_{OUT} 10% to 90%	$A_V = 1$, $V_{OUT} = 100mV_{P-P}$, $R_f = 0\Omega$, $R_L = 2k\Omega$ to V_{CM}		264		ns
	Fall Time, V_{OUT} 90% to 10%	$A_V = 1$, $V_{OUT} = 100mV_{P-P}$, $R_f = 0\Omega$, $R_L = 2k\Omega$ to V_{CM}		254		ns
t_s	Settling Time to 0.01% 10V Step; 10% to V_{OUT}	$A_V = -1$, $V_{OUT} = 10V_{P-P}$, $R_g = R_f = 10k$, $R_L = 2k\Omega$ to V_{CM}		27		μs

Electrical Specifications, $V_S \pm 5V$ $V_{CM} = 0$, $V_O = 0V$, $T_A = +25^\circ C$, unless otherwise noted. **Boldface entries apply over the operating temperature range, $-40^\circ C$ to $+125^\circ C$. Temperature data established by characterization.**

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 8)	TYP	MAX (Note 8)	UNIT
V_{OS}	Input Offset Voltage	ISL28208	-230	25	230	μV
		ISL28408	-330		330	μV
		ISL28108 SOIC, TDFN	-150	10	150	μV
			-270		270	μV
TCV_{OS}	Input Offset Voltage Temperature Coefficient	ISL28208 SOIC $-40^\circ C$ to $+125^\circ C$		0.1	1.1	$\mu V/^\circ C$
		ISL28208 MSOP $-40^\circ C$ to $+125^\circ C$		0.2	1.5	$\mu V/^\circ C$
		ISL28208 TDFN ISL28408 $-40^\circ C$ to $+125^\circ C$		0.2	1.4	$\mu V/^\circ C$
		ISL28108 SOIC, TDFN $-40^\circ C$ to $+125^\circ C$		0.2	1.2	$\mu V/^\circ C$
ΔV_{OS}	Input Offset Voltage Match (ISL28208 only)		-300	3	300	μV
			-400		400	μV
I_B	Input Bias Current		-43	-15		nA
			-63			nA
TCI_B	Input Bias Current Temperature Coefficient	$-40^\circ C$ to $+125^\circ C$		-0.067		nA/ $^\circ C$
I_{OS}	Input Offset Current	ISL28208	-3	0	3	nA
			-4		4	nA
		ISL28108 SOIC, TDFN ISL28408	-4	0	4	nA
			-5		5	nA
CMRR	Common-Mode Rejection Ratio	$V_{CM} = V_- - 0.5V$ to $V_+ - 1.8V$		101		dB
		$V_{CM} = V_- - 0.2V$ to $V_+ - 1.8V$		123		dB
				89		dB
		$V_{CM} = V_-$ to $V_+ - 1.8V$ ISL28108, ISL28208	105	123		dB
			100	112		dB
		$V_{CM} = V_-$ to $V_+ - 1.8V$ ISL28408	105	123		dB
			97	112		dB
V_{CMIR}	Common Mode Input Voltage Range	Guaranteed by CMRR test	$V_- - 0.5$		$V_+ - 1.8$	V
			V_-		$V_+ - 1.8$	V
PSRR	Power Supply Rejection Ratio	$V_S = 3V$ to $10V$, V_{CMIR} = Valid Input Voltage All except ISL28208 MSOP	110	126		dB
			109	123		dB
		ISL28208 MSOP $V_S = 3V$ to $10V$, V_{CMIR} = Valid Input Voltage	109	126		dB
			107	123		dB
A_{VOL}	Open-Loop Gain	$V_O = -3V$ to $+3V$, $R_L = 10k\Omega$ to ground	117	124		dB
			99			dB
V_{OL}	Output Voltage Low, V_{OUT} to V_-	$R_L = 10k\Omega$		23	38	mV
					48	mV
V_{OH}	Output Voltage High, V_+ to V_{OUT}	$R_L = 10k\Omega$		30	65	mV
					70	mV

Electrical Specifications, $V_S \pm 5V$ $V_{CM} = 0$, $V_O = 0V$, $T_A = +25^\circ C$, unless otherwise noted. **Boldface entries apply over the operating temperature range, $-40^\circ C$ to $+125^\circ C$. Temperature data established by characterization. (Continued)**

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 8)	TYP	MAX (Note 8)	UNIT
I_S	Supply Current/Amplifier	$R_L = \text{Open}$		165	250	μA
				240	350	μA
I_{SC+}	Output Short Circuit Source Current	$R_L = 10\Omega$ to V_-		14		mA
I_{SC-}	Output Short Circuit Sink Current	$R_L = 10\Omega$ to V_+		22		mA
AC SPECIFICATIONS						
GBW	Gain Bandwidth Product	$A_{CL} = 101$, $V_O = 100mV_{P-P}$, $R_L = 2k\Omega$		1.2		MHz
e_{np-p}	Noise Voltage	0.1Hz to 10Hz		600		nV _{P-P}
e_n	Noise Voltage Density	$f = 10\text{Hz}$		18		nV/ $\sqrt{\text{Hz}}$
e_n	Noise Voltage Density	$f = 100\text{Hz}$		16		nV/ $\sqrt{\text{Hz}}$
e_n	Noise Voltage Density	$f = 1\text{kHz}$		15.8		nV/ $\sqrt{\text{Hz}}$
e_n	Noise Voltage Density	$f = 10\text{kHz}$		15.8		nV/ $\sqrt{\text{Hz}}$
i_n	Noise Current Density	$f = 10\text{kHz}$		90		fA/ $\sqrt{\text{Hz}}$
TRANSIENT RESPONSE						
SR	Slew Rate, V_{OUT} 20% to 80%	$A_V = 1$, $R_L = 2k\Omega$, $V_O = 4V_{P-P}$		0.4		V/ μs
t_r , t_f , Small Signal	Rise Time, V_{OUT} 10% to 90%	$A_V = 1$, $V_{OUT} = 100mV_{P-P}$, $R_f = 0\Omega$, $R_L = 2k\Omega$ to V_{CM}		264		ns
	Fall Time, V_{OUT} 90% to 10%	$A_V = 1$, $V_{OUT} = 100mV_{P-P}$, $R_f = 0\Omega$, $R_L = 2k\Omega$ to V_{CM}		254		ns
t_s	Settling Time to 0.01% 4V Step; 10% to V_{OUT}	$A_V = -1$, $V_{OUT} = 4V_{P-P}$, $R_g = R_f = 10k$, $R_L = 2k\Omega$ to V_{CM}		14.4		μs

NOTE:

8. Compliance to data sheet limits is assured by one or more methods: production test, characterization and/or design.

Typical Performance Curves $V_S = \pm 15V$, $V_{CM} = 0V$, $R_L = \text{Open}$, $T_A = +25^\circ C$, unless otherwise specified.

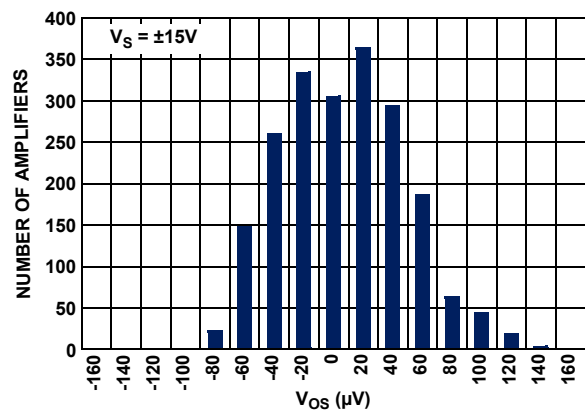


FIGURE 3. ISL28408 SOIC INPUT OFFSET VOLTAGE DISTRIBUTION, $V_S = \pm 15V$

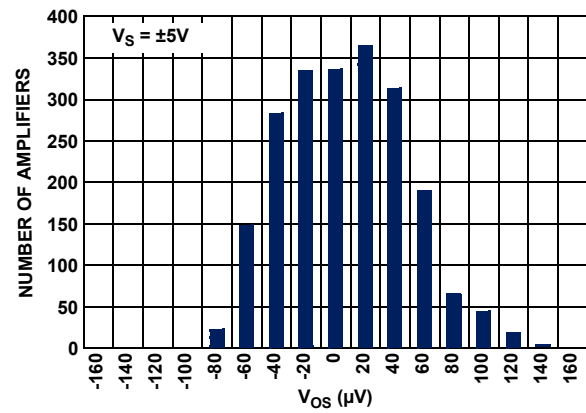


FIGURE 4. ISL28408 SOIC INPUT OFFSET VOLTAGE DISTRIBUTION, $V_S = \pm 5V$

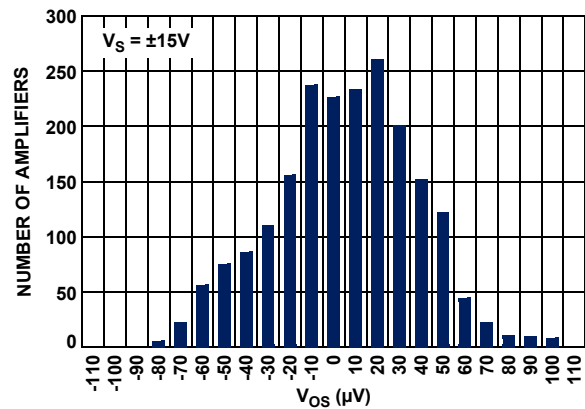


FIGURE 5. ISL28208 INPUT OFFSET VOLTAGE DISTRIBUTION, $V_S = \pm 15V$

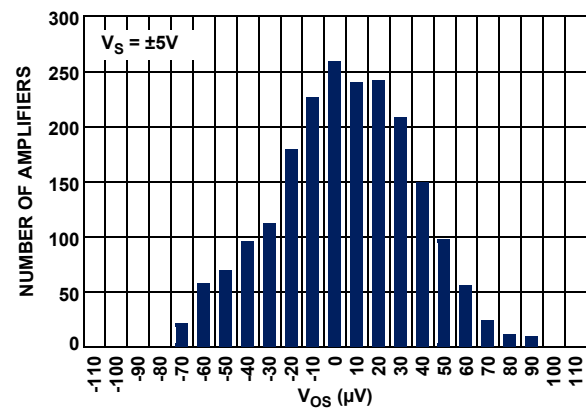


FIGURE 6. ISL28208 INPUT OFFSET VOLTAGE DISTRIBUTION, $V_S = \pm 5V$

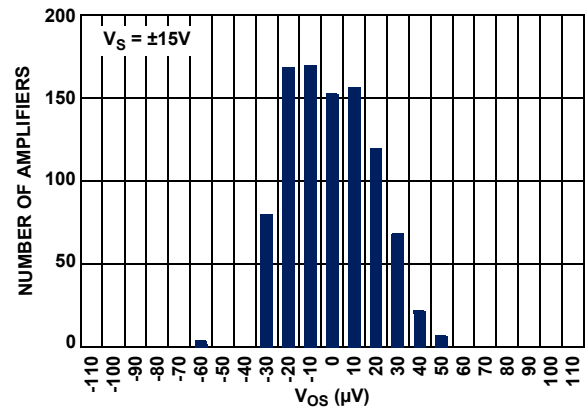


FIGURE 7. ISL28108 SOIC INPUT OFFSET VOLTAGE DISTRIBUTION, $V_S = \pm 15V$

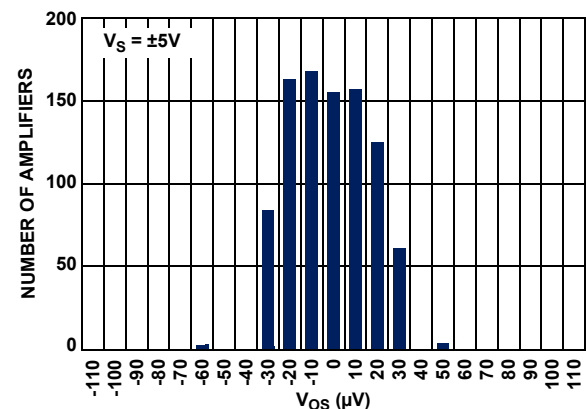


FIGURE 8. ISL28108 SOIC INPUT OFFSET VOLTAGE DISTRIBUTION, $V_S = \pm 5V$

Typical Performance Curves $V_S = \pm 15V, V_{CM} = 0V, R_L = \text{Open}, T_A = +25^\circ C$, unless otherwise specified. (Continued)

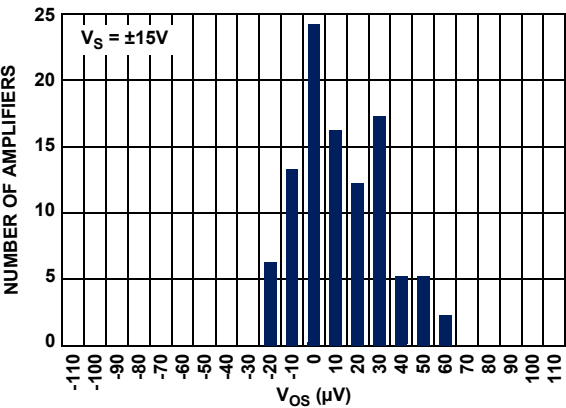


FIGURE 9. ISL28108 TDFN INPUT OFFSET VOLTAGE DISTRIBUTION, $V_S = \pm 15V$

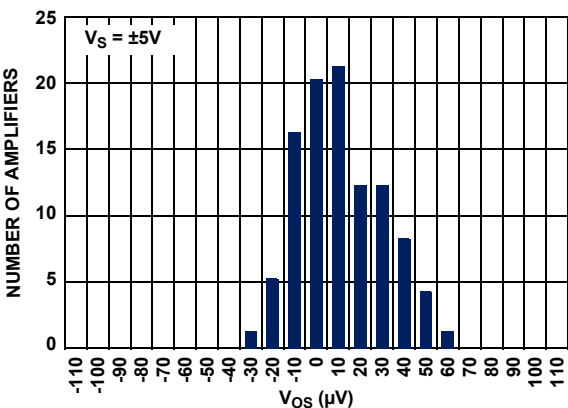


FIGURE 10. ISL28108 TDFN INPUT OFFSET VOLTAGE DISTRIBUTION, $V_S = \pm 5V$

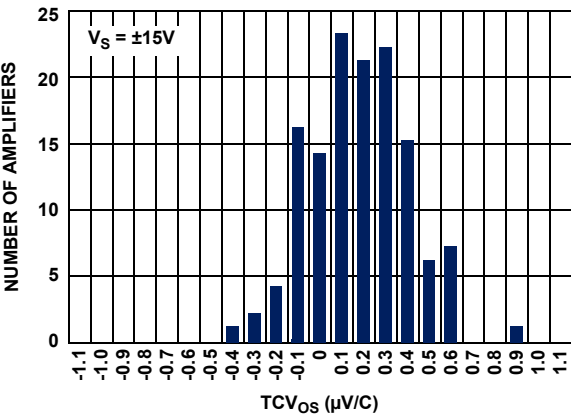


FIGURE 11. ISL28408 SOIC TCV_{OS} vs NUMBER OF AMPLIFIERS, $V_S = \pm 15V$

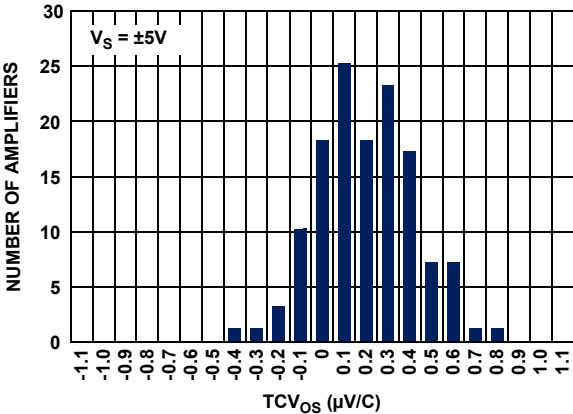


FIGURE 12. ISL28408 SOIC TCV_{OS} vs NUMBER OF AMPLIFIERS, $V_S = \pm 5V$

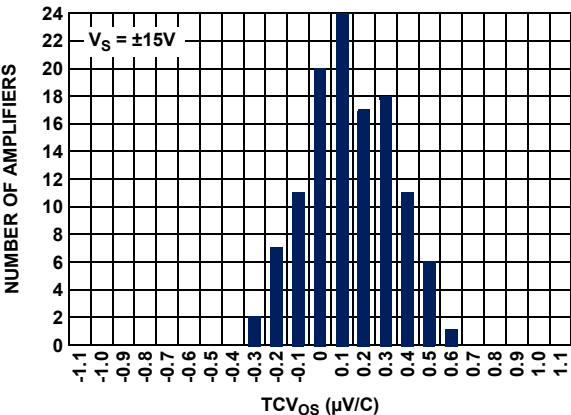


FIGURE 13. ISL28208 SOIC TCV_{OS} vs NUMBER OF AMPLIFIERS, $V_S = \pm 15V$

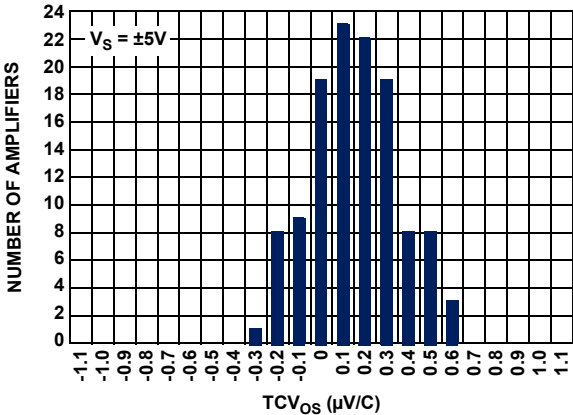


FIGURE 14. ISL28208 SOIC TCV_{OS} vs NUMBER OF AMPLIFIERS, $V_S = \pm 5V$

Typical Performance Curves

$V_S = \pm 15V$, $V_{CM} = 0V$, $R_L = \text{Open}$, $T_A = +25^\circ\text{C}$, unless otherwise specified. (Continued)

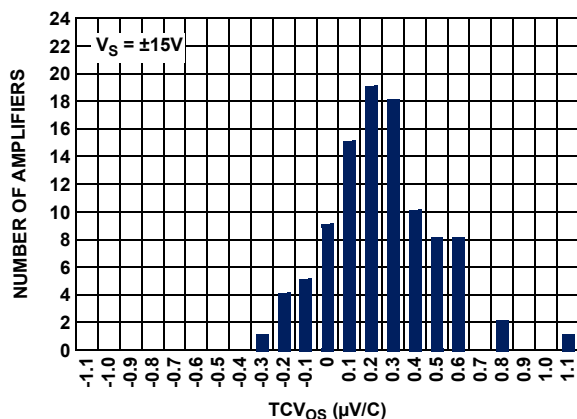


FIGURE 15. ISL28208 TDFN AND MSOP TCV_{OS} vs NUMBER OF AMPLIFIERS, $V_S = \pm 15V$

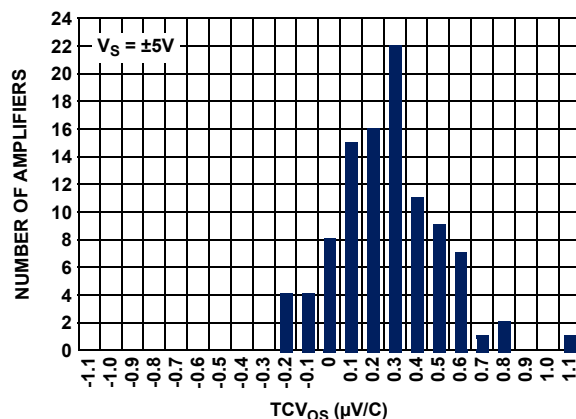


FIGURE 16. ISL28208 TDFN AND MSOP TCV_{OS} vs NUMBER OF AMPLIFIERS, $V_S = \pm 5V$

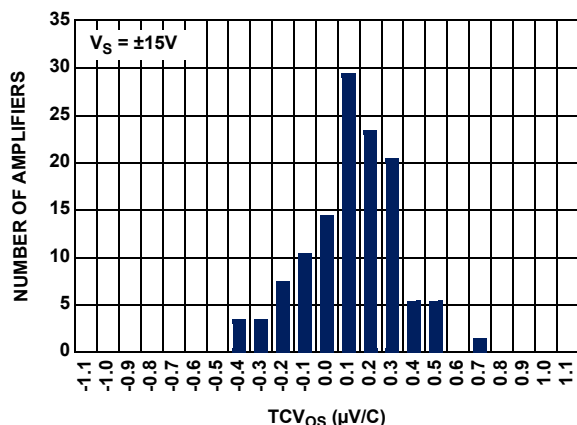


FIGURE 17. ISL28108 SOIC TCV_{OS} vs NUMBER OF AMPLIFIERS, $V_S = \pm 15V$

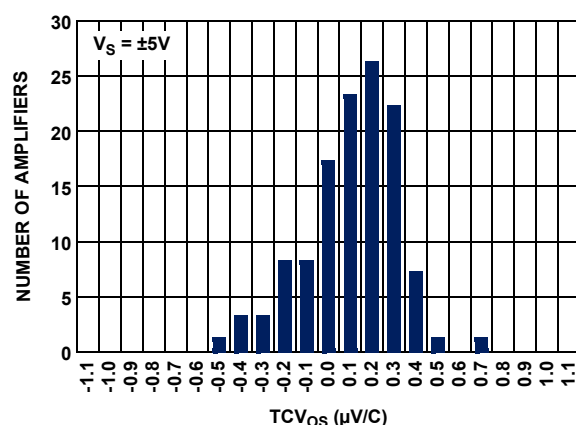


FIGURE 18. ISL28108 SOIC TCV_{OS} vs NUMBER OF AMPLIFIERS, $V_S = \pm 5V$

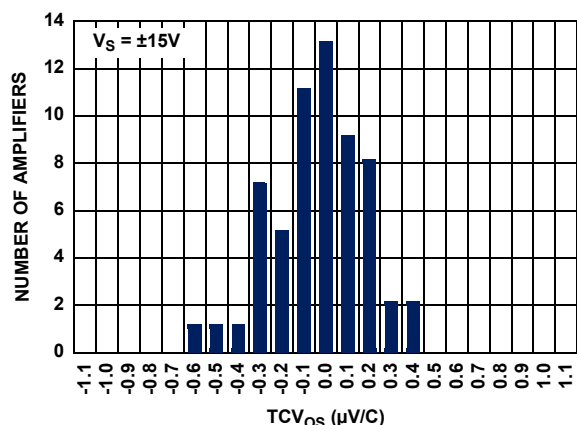


FIGURE 19. ISL28108 TDFN TCV_{OS} vs NUMBER OF AMPLIFIERS, $V_S = \pm 15V$

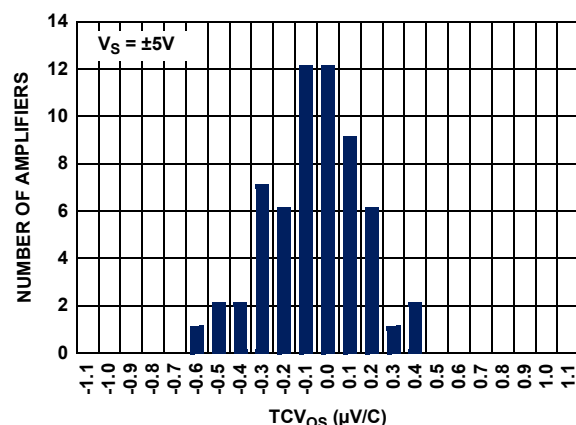


FIGURE 20. ISL28108 TDFN TCV_{OS} vs NUMBER OF AMPLIFIERS, $V_S = \pm 5V$

Typical Performance Curves $V_S = \pm 15V$, $V_{CM} = 0V$, $R_L = \text{Open}$, $T_A = +25^\circ C$, unless otherwise specified. (Continued)

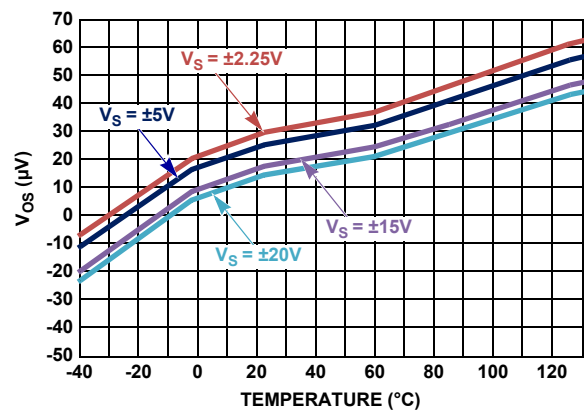


FIGURE 21. V_{OS} vs TEMPERATURE

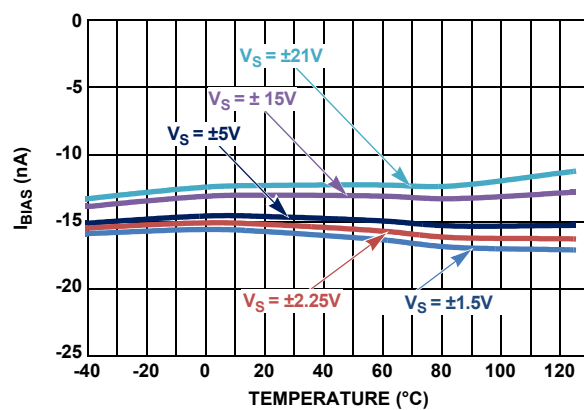


FIGURE 22. I_{BIAS} vs TEMPERATURE vs SUPPLY

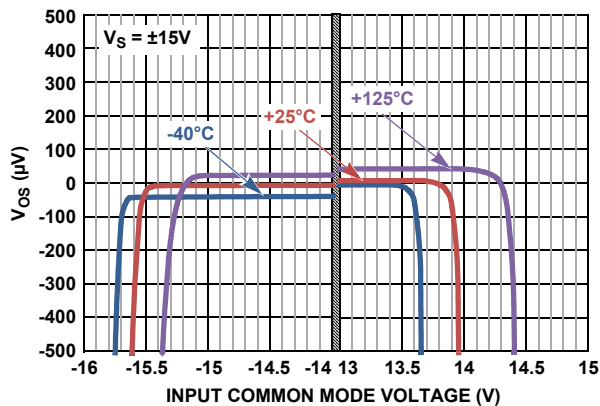


FIGURE 23. INPUT OFFSET VOLTAGE vs INPUT COMMON MODE VOLTAGE, $V_S = \pm 15V$

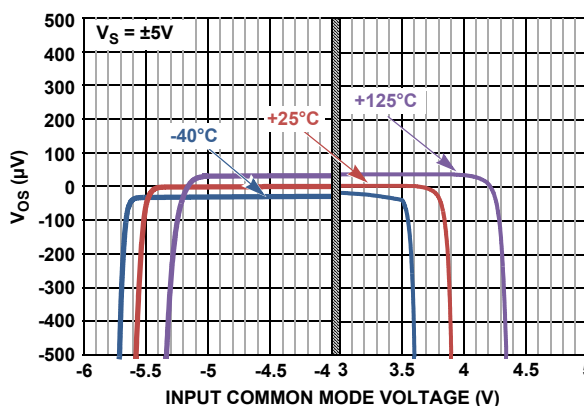


FIGURE 24. INPUT OFFSET VOLTAGE vs INPUT COMMON MODE VOLTAGE, $V_S = \pm 5V$

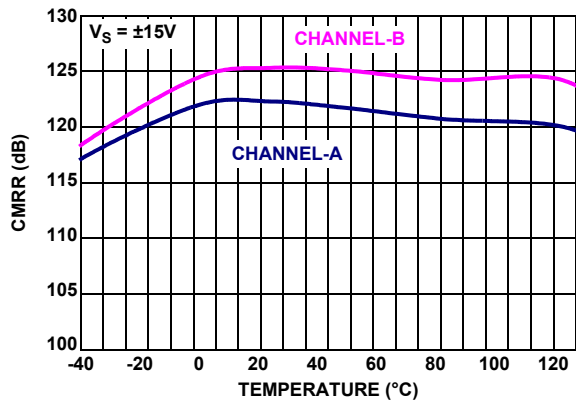


FIGURE 25. CMRR vs TEMPERATURE, $V_S = \pm 15V$

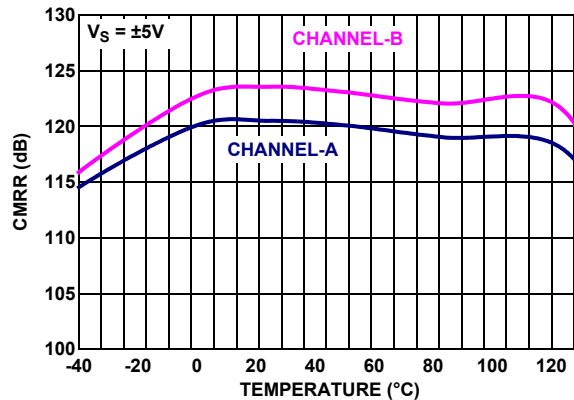


FIGURE 26. CMRR vs TEMPERATURE, $V_S = \pm 5V$

Typical Performance Curves $V_S = \pm 15V, V_{CM} = 0V, R_L = \text{Open}, T_A = +25^\circ C$, unless otherwise specified. (Continued)

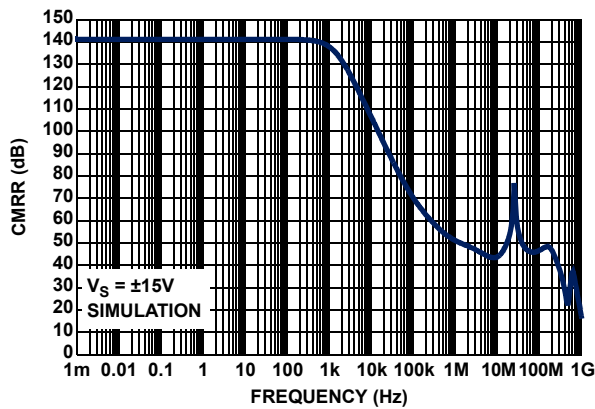


FIGURE 27. CMRR vs FREQUENCY, $V_S = \pm 15V$

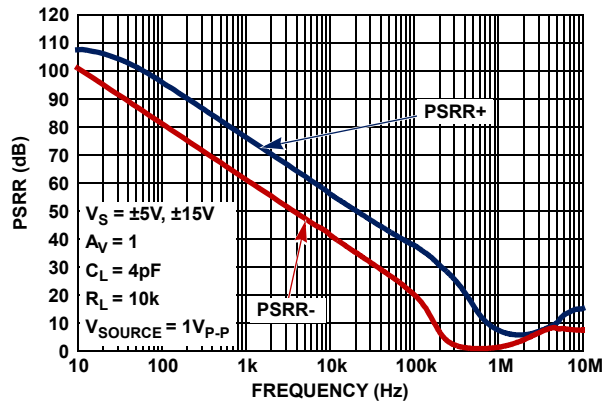


FIGURE 28. PSRR vs FREQUENCY, $V_S = \pm 5V$ AND $\pm 15V$

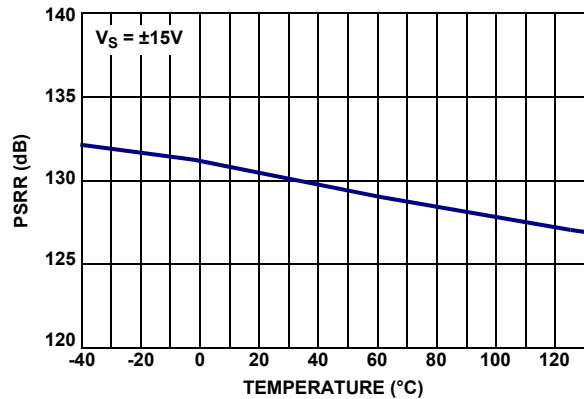


FIGURE 29. PSRR (DC) vs TEMPERATURE, $V_S = \pm 15V$

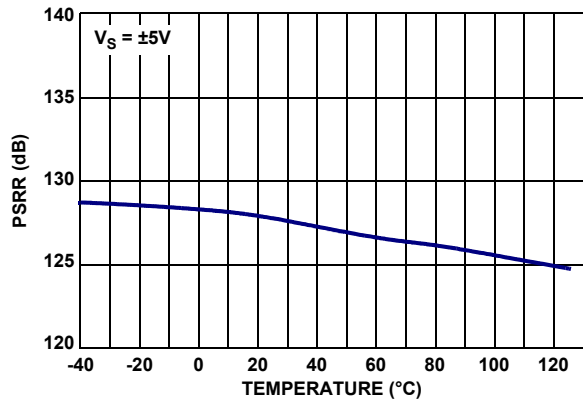


FIGURE 30. PSRR (DC) vs TEMPERATURE, $V_S = \pm 5V$

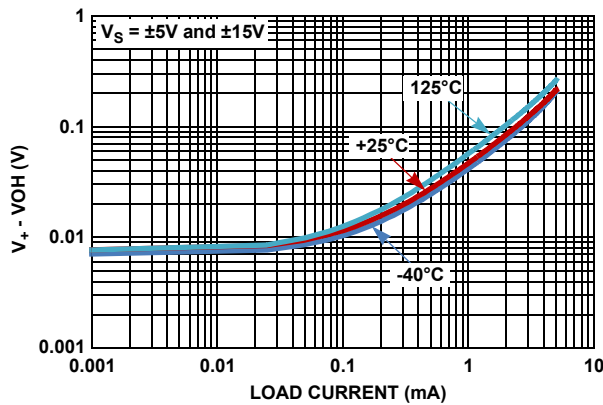


FIGURE 31. OUTPUT OVERHEAD VOLTAGE HIGH vs LOAD CURRENT, $V_S = \pm 5V$ AND $\pm 15V$

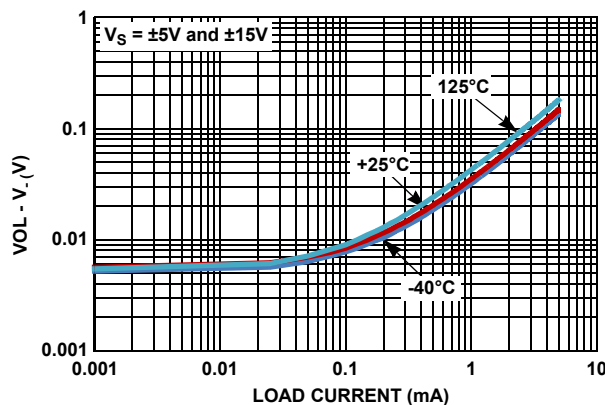


FIGURE 32. OUTPUT OVERHEAD VOLTAGE LOW vs LOAD CURRENT, $V_S = \pm 5V$ AND $\pm 15V$

Typical Performance Curves $V_S = \pm 15V, V_{CM} = 0V, R_L = \text{Open}, T_A = +25^\circ\text{C}$, unless otherwise specified. (Continued)

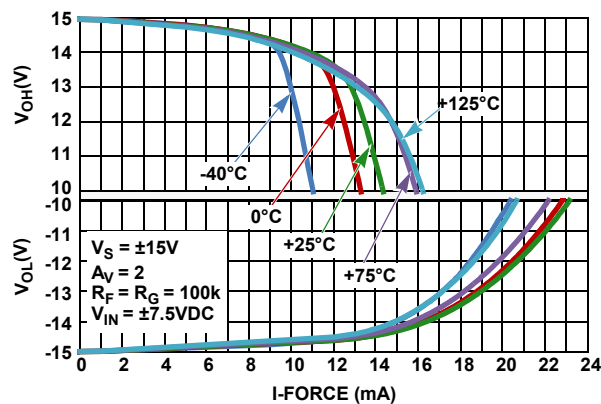


FIGURE 33. ISL28208 OUTPUT VOLTAGE SWING vs LOAD CURRENT
 $V_S = \pm 15V$

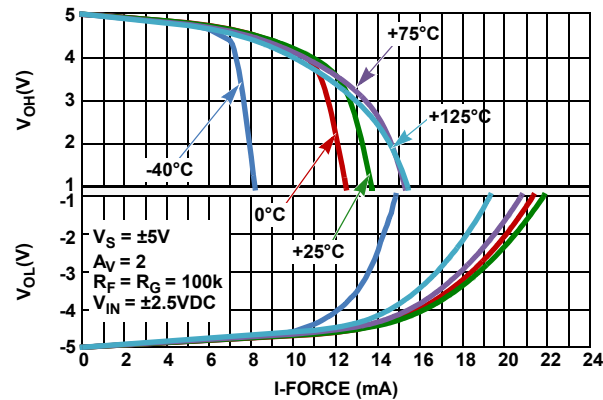


FIGURE 34. ISL28208 OUTPUT VOLTAGE SWING vs LOAD CURRENT
 $V_S = \pm 5V$

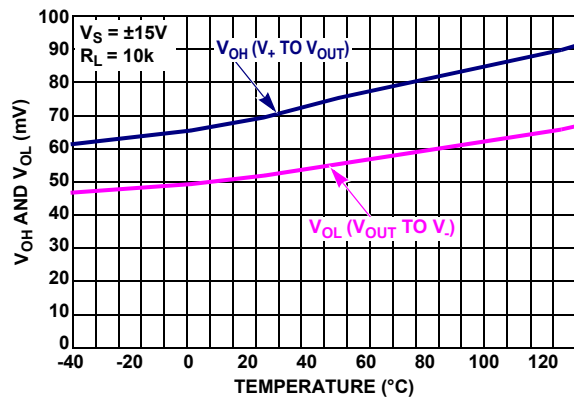


FIGURE 35. V_{OUT} HIGH AND LOW vs TEMPERATURE,
 $V_S = \pm 15V, R_L = 10k$

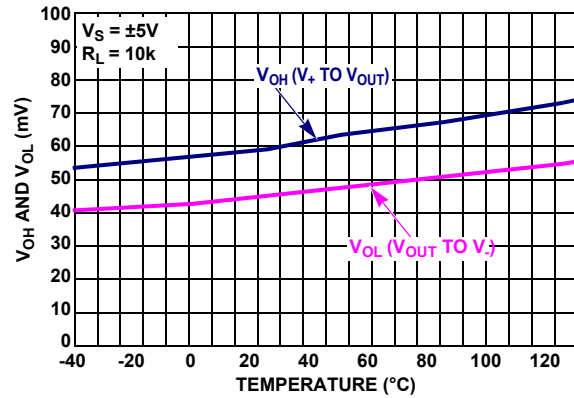


FIGURE 36. V_{OUT} HIGH AND LOW vs TEMPERATURE,
 $V_S = \pm 5V, R_L = 10k$

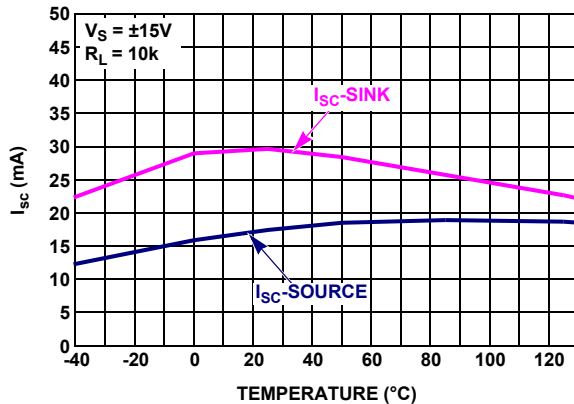


FIGURE 37. SHORT CIRCUIT CURRENT vs TEMPERATURE, $V_S = \pm 15V$

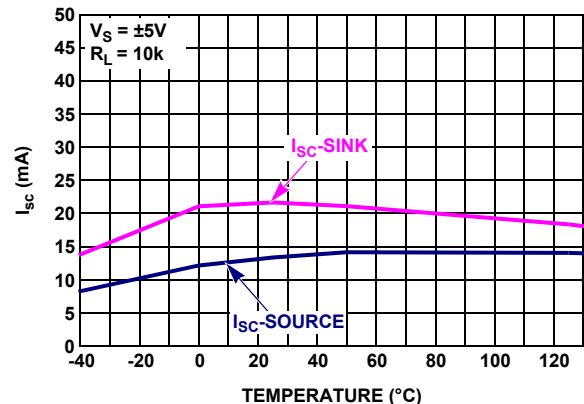


FIGURE 38. SHORT CIRCUIT CURRENT vs TEMPERATURE, $V_S = \pm 5V$

Typical Performance Curves $V_S = \pm 15V, V_{CM} = 0V, R_L = \text{Open}, T_A = +25^\circ C$, unless otherwise specified. (Continued)

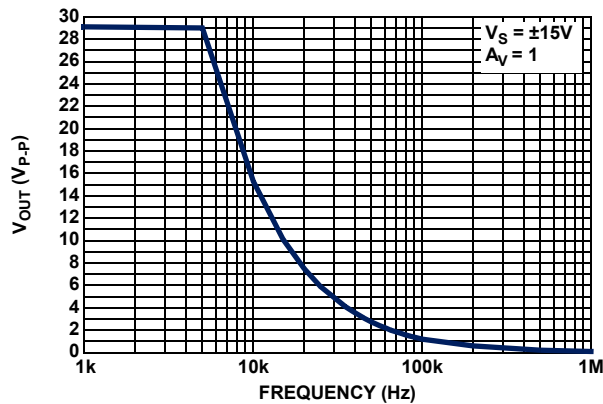


FIGURE 39. MAX OUTPUT VOLTAGE vs FREQUENCY

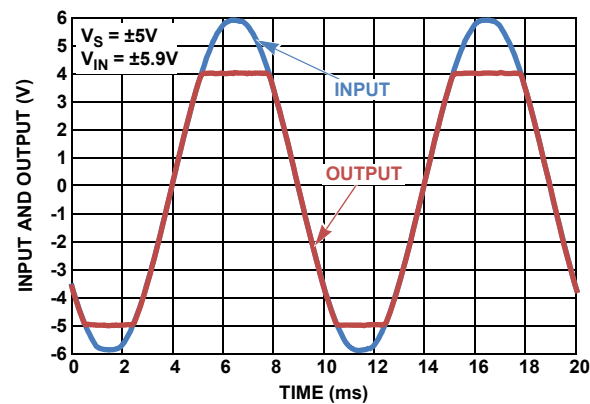


FIGURE 40. NO PHASE REVERSAL

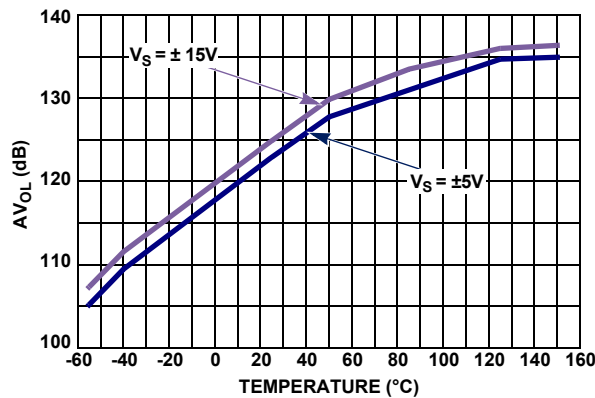


FIGURE 41. AV_{OL} vs TEMPERATURE

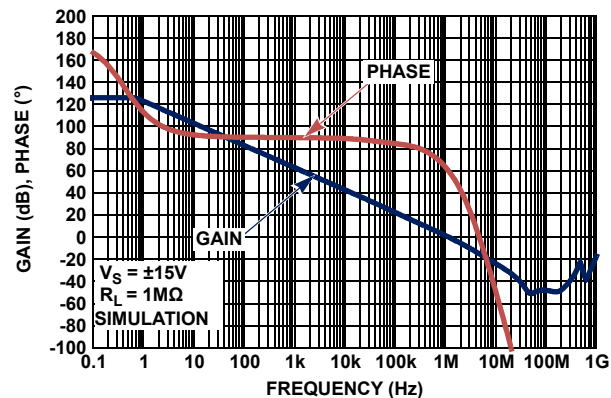


FIGURE 42. OPEN-LOOP GAIN, PHASE vs FREQUENCY, $V_S = \pm 15V$

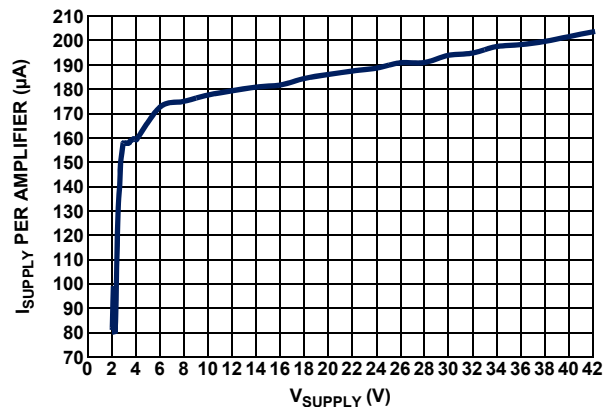


FIGURE 43. SUPPLY CURRENT vs SUPPLY VOLTAGE

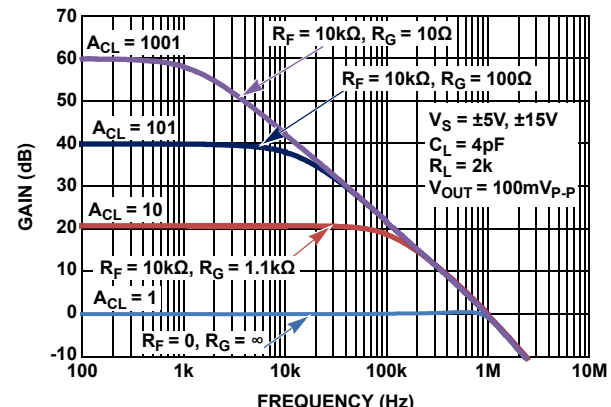


FIGURE 44. FREQUENCY RESPONSE vs CLOSED LOOP GAIN

Typical Performance Curves $V_S = \pm 15V, V_{CM} = 0V, R_L = \text{Open}, T_A = +25^\circ C$, unless otherwise specified. (Continued)

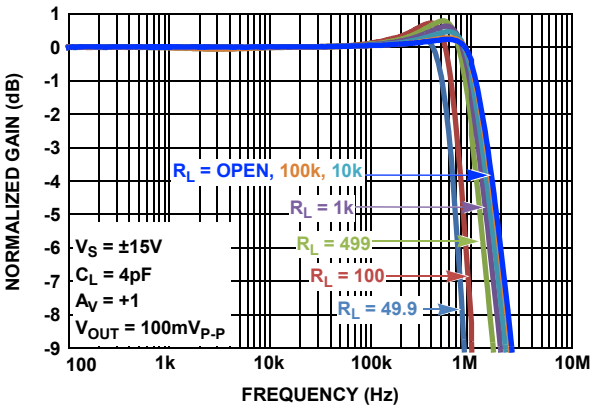


FIGURE 45. GAIN vs FREQUENCY vs $R_L, V_S = \pm 15V$

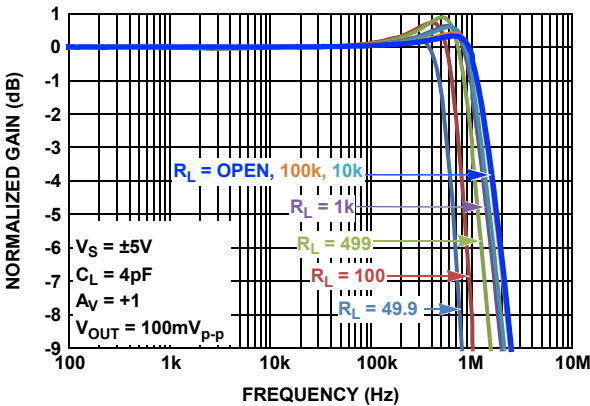


FIGURE 46. GAIN vs FREQUENCY vs $R_L, V_S = \pm 5V$

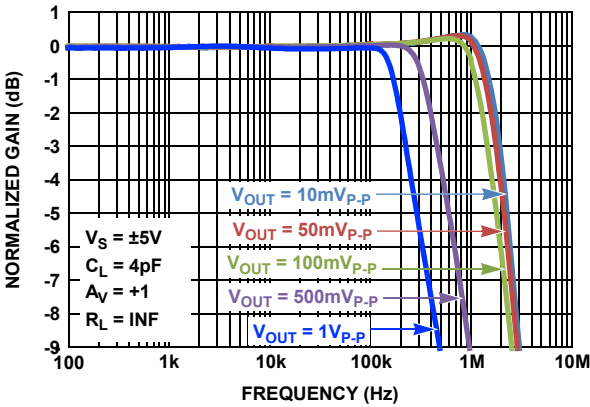


FIGURE 47. GAIN vs FREQUENCY vs OUTPUT VOLTAGE

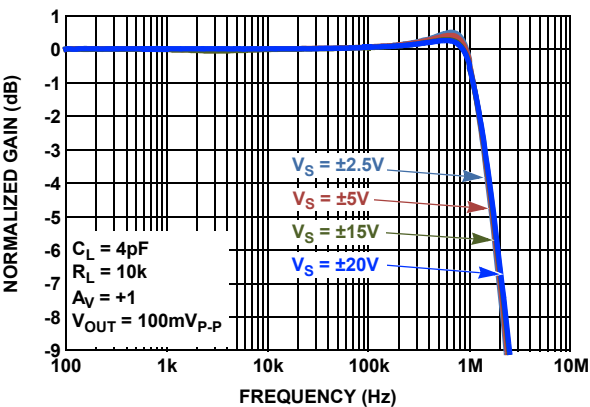


FIGURE 48. GAIN vs FREQUENCY vs SUPPLY VOLTAGE

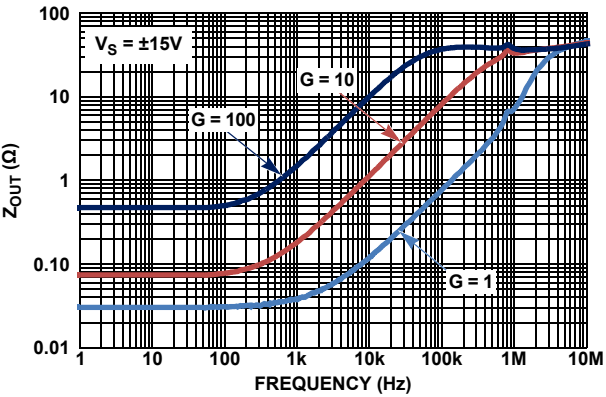


FIGURE 49. OUTPUT IMPEDANCE vs FREQUENCY, $V_S = \pm 15V$

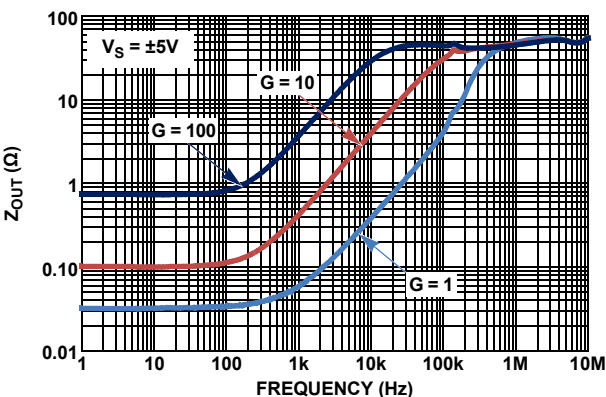


FIGURE 50. OUTPUT IMPEDANCE vs FREQUENCY, $V_S = \pm 5V$

Typical Performance Curves $V_S = \pm 15V, V_{CM} = 0V, R_L = \text{Open}, T_A = +25^\circ C$, unless otherwise specified. (Continued)

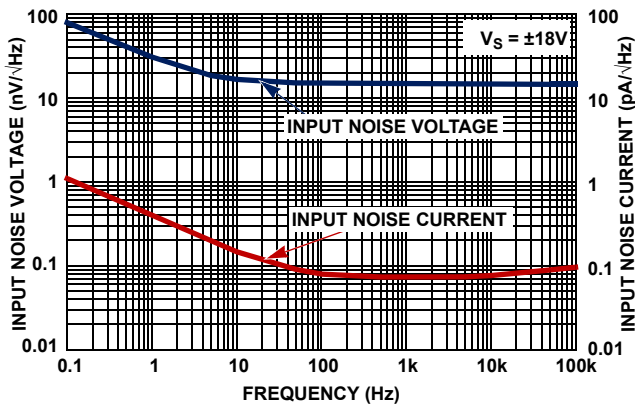


FIGURE 51. INPUT NOISE VOLTAGE (en) AND CURRENT (in) vs FREQUENCY, $V_S = \pm 18V$

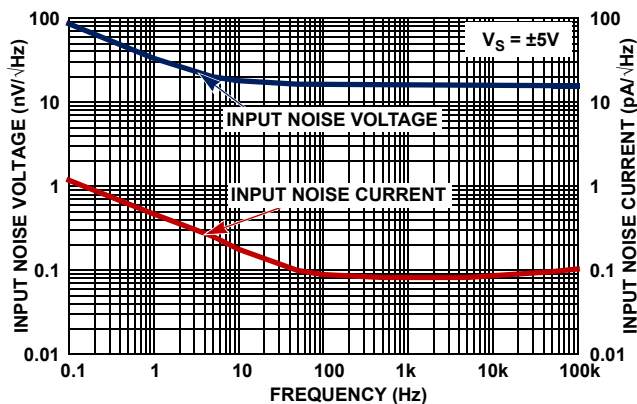


FIGURE 52. INPUT NOISE VOLTAGE (en) AND CURRENT (in) vs FREQUENCY, $V_S = \pm 5V$

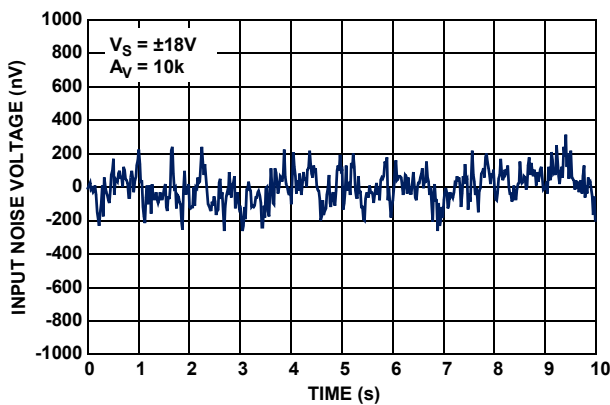


FIGURE 53. INPUT NOISE VOLTAGE 0.1Hz TO 10Hz, $V_S = \pm 18V$

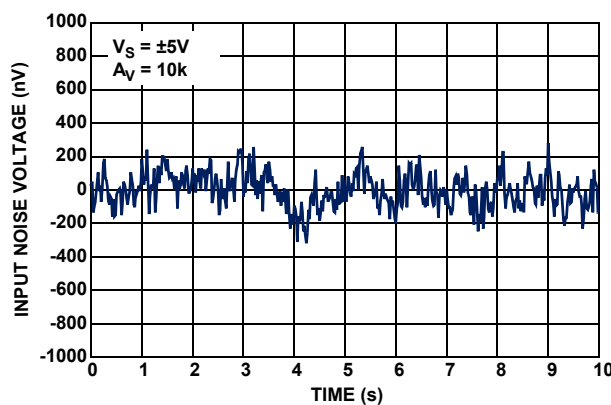


FIGURE 54. INPUT NOISE VOLTAGE 0.1Hz TO 10Hz, $V_S = \pm 5V$

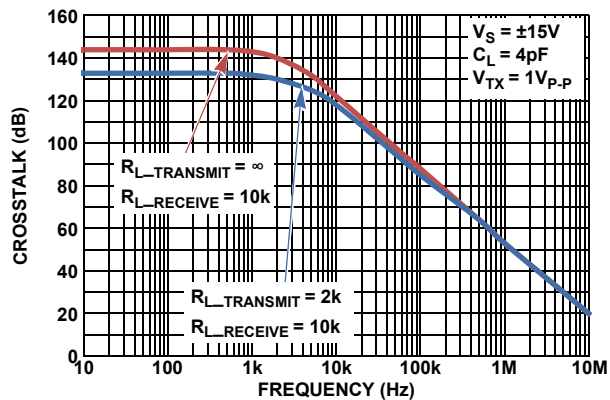


FIGURE 55. ISL28208 CHANNEL SEPARATION vs FREQUENCY, $V_S = \pm 5V, \pm 15V$

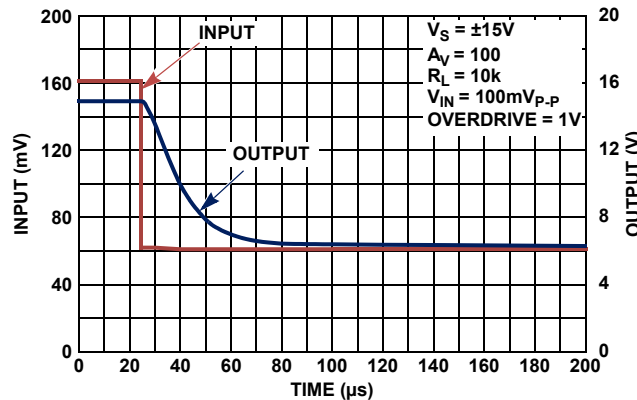


FIGURE 56. POSITIVE OUTPUT OVERLOAD RESPONSE TIME, $V_S = \pm 15V$

Typical Performance Curves $V_S = \pm 15V, V_{CM} = 0V, R_L = \text{Open}, T_A = +25^\circ C$, unless otherwise specified. (Continued)

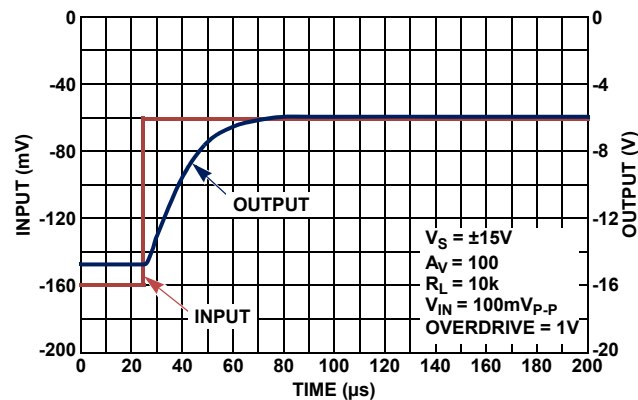


FIGURE 57. NEGATIVE OUTPUT OVERLOAD RESPONSE TIME, $V_S = \pm 15V$

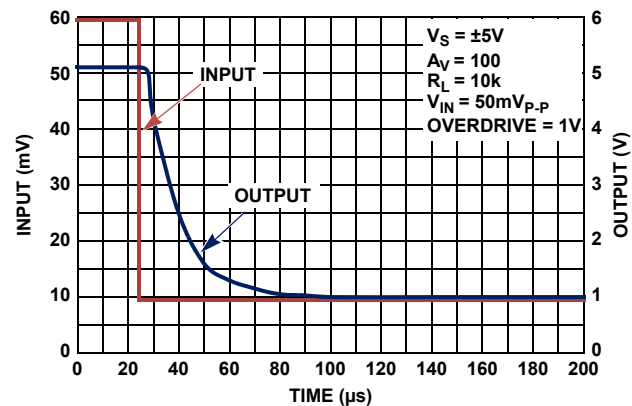


FIGURE 58. POSITIVE OUTPUT OVERLOAD RESPONSE TIME, $V_S = \pm 5V$

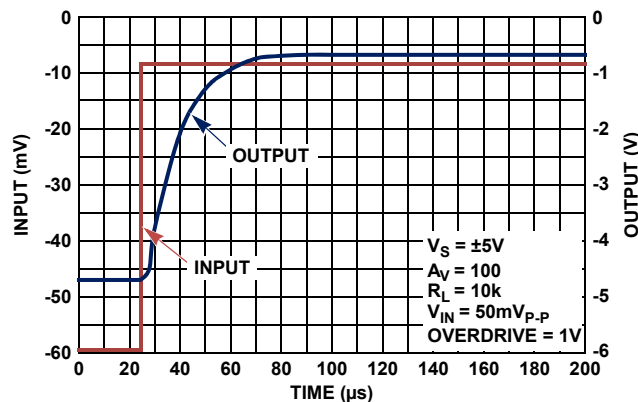


FIGURE 59. NEGATIVE OUTPUT OVERLOAD RESPONSE TIME, $V_S = \pm 5V$

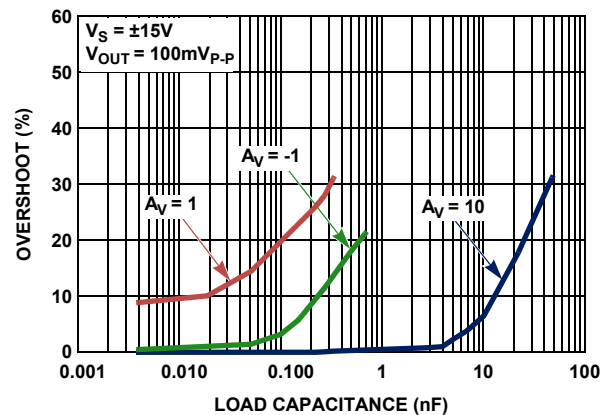


FIGURE 60. OVERSHOOT vs CAPACITIVE LOAD, $V_S = \pm 15V$

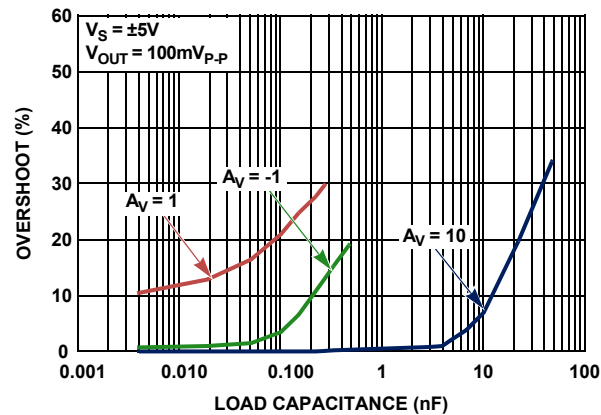


FIGURE 61. OVERSHOOT vs CAPACITIVE LOAD, $V_S = \pm 5V$

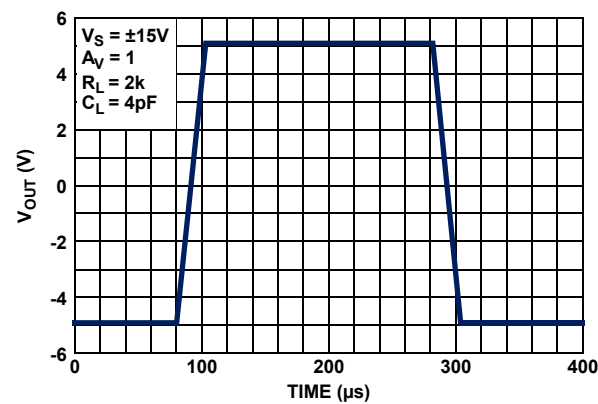


FIGURE 62. LARGE SIGNAL 10V STEP RESPONSE, $V_S = \pm 15V$

Typical Performance Curves $V_S = \pm 15V, V_{CM} = 0V, R_L = \text{Open}, T_A = +25^\circ C$, unless otherwise specified. (Continued)

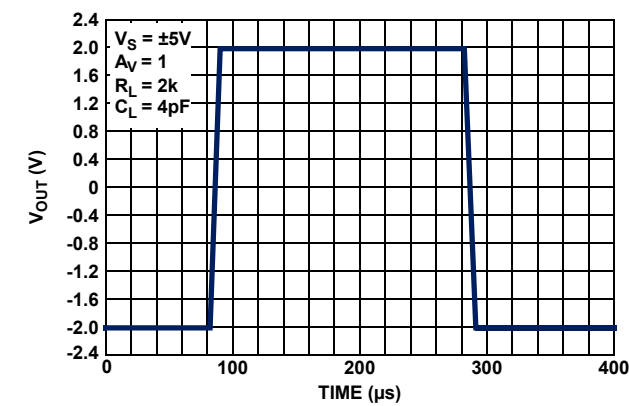


FIGURE 63. LARGE SIGNAL 4V STEP RESPONSE, $V_S = \pm 5V$

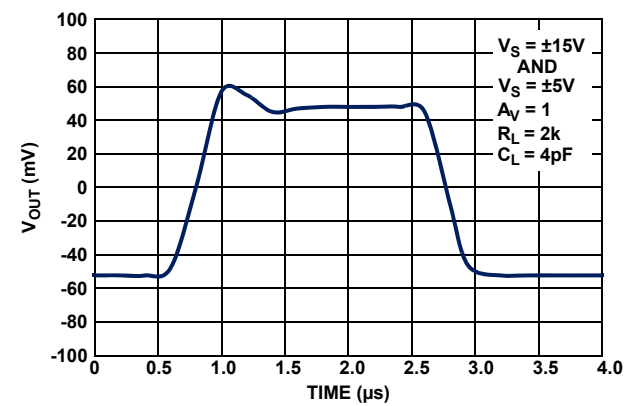


FIGURE 64. SMALL SIGNAL TRANSIENT RESPONSE $V_S = \pm 5V, \pm 15V$

Applications Information

Functional Description

The ISL28108, ISL28208, and ISL28408 are single, dual and quad, 1.2MHz, single supply rail-to-rail output amplifiers with a common mode input voltage range extending to a range of 0.5V below the V- rail. Their input stages are optimized for precision sensing of ground referenced signals in low voltage, single supply applications. The input stage has the capability of handling large input differential voltages without phase inversion making them suitable for high voltage comparator applications. Their bipolar design features high open loop gain and excellent DC input and output temperature stability. These op amps feature low quiescent current of 165µA, and a maximum temperature drift ranging from 1.1µV/°C for the ISL28208 and ISL28408 in the SOIC package to 1.4µV/°C for the ISL28208 in the TDFN package and the ISL28408 in the SOIC package (see Figures 11 through 20). All devices are fabricated in a new precision 40V complementary bipolar DI process and immune from latch-up.

Operating Voltage Range

The devices are designed to operate over the 3V (±1.5V) to 40V (±20V) range and are fully characterized at ±5V and ±15V. Both DC and AC performance remain virtually unchanged over the ±5V to ±15V operating voltage range. Parameter variation with operating voltage is shown in the "Typical Performance Curves" beginning on page 10.

Input Stage Performance

The PNP input stage has a common mode input range extending up to 0.5V below ground at +25°C (see Figures 23 and 24). Full amplifier performance is guaranteed down to ground (V-) over the -40°C to +125°C temperature range. For common mode voltages down to -0.5V the amplifiers are fully functional, but performance degrades slightly over the full temperature range. This feature provides excellent CMRR, AC performance and DC accuracy when amplifying low level ground referenced signals.

The input stage has a maximum input differential voltage equal to a diode drop greater than the supply voltage (max 42V) and does not contain the back-to-back input protection diodes found on many similar amplifiers. This feature enables the device to function as a precision comparator by maintaining very high input impedance for high voltage differential input comparator voltages. The high differential input impedance also enables the device to operate reliably in large signal pulse applications without the need for anti-parallel clamp diodes required on MOSFET and most bipolar input stage op amps. Thus, input signal distortion caused by nonlinear clamps under high slew rate conditions are avoided.

In applications where one or both amplifier input terminals are at risk of exposure to voltages beyond the supply rails, current limiting resistors may be needed at each input terminal (see Figure 65, R_{IN+} , R_{IN-}) to limit current through the power supply ESD diodes to 20mA.

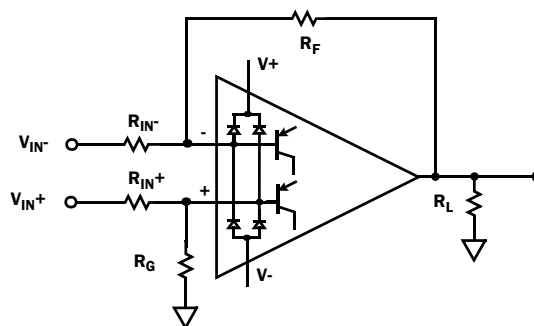


FIGURE 65. INPUT ESD DIODE CURRENT LIMITING

Output Drive Capability

The bipolar rail-to-rail output stage features low saturation levels that enable an output voltage swing to less than 10mV when the total output load (including feedback resistance) is held below 50µA (Figures 31 and 32). With ±15V supplies this can be achieved by using feedback resistor values >300kΩ. The low input bias and offset currents (-43nA and ±3nA +25°C max respectively) minimize DC offset errors at these high resistance values. For example, a balanced 4 resistor gain circuit (Figure 65) with 1MΩ feedback resistors (R_F , R_G) generates a worst case input offset error of only ±3mV. Furthermore, the low noise current reduces the added noise associated with high feedback resistance.

The output stage is internally current limited. Output current limit over-temperature is shown in Figures 37 and 38. The amplifiers can withstand a short circuit to either rail as long as the power dissipation limits are not exceeded. This applies to only one amplifier at a time for the dual op amp. Continuous operation under these conditions may degrade long-term reliability.

The amplifiers perform well driving capacitive loads (Figures 60 and 61). The unity gain, voltage follower (buffer) configuration provides the highest bandwidth, but is also the most sensitive to ringing produced by load capacitance found in BNC cables. Unity gain overshoot is limited to 30% at capacitance values to 0.33nF. At gains of 10 and higher, the device is capable of driving more than 10nF without significant overshoot.

Output Phase Reversal

Output phase reversal is a change of polarity in the amplifier transfer function when the input voltage exceeds the supply voltage. These devices are immune to output phase reversal, out to 0.5V beyond the rail ($V_{ABS\ MAX}$) limit (see Figure 40).

Unused Channels

If the application requires only one channel, the user must configure any unused channel to prevent it from oscillating. Unused channels can oscillate if the input and output pins are floating. This will result in higher-than-expected supply currents and possible noise injection into the channel being used. The proper way to prevent oscillation is to short the output to the inverting input, and ground the positive input (Figure 66).

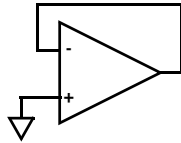


FIGURE 66. PREVENTING OSCILLATIONS IN UNUSED CHANNELS

Power Dissipation

It is possible to exceed the +150°C maximum junction temperatures under certain load and power supply conditions. It is therefore important to calculate the maximum junction temperature (T_{JMAX}) for all applications to determine if power supply voltages, load conditions, or package type need to be modified to remain in the safe operating area. These parameters are related using Equation 1:

$$T_{JMAX} = T_{MAX} + \theta_{JA} \times PD_{MAXTOTAL} \quad (EQ. 1)$$

where:

- $PD_{MAXTOTAL}$ is the sum of the maximum power dissipation of each amplifier in the package (PD_{MAX})
- PD_{MAX} for each amplifier can be calculated using Equation 2:

$$PD_{MAX} = V_S \times I_{qMAX} + (V_S - V_{OUTMAX}) \times \frac{V_{OUTMAX}}{R_L} \quad (EQ. 2)$$

where:

- T_{MAX} = Maximum ambient temperature
- θ_{JA} = Thermal resistance of the package
- PD_{MAX} = Maximum power dissipation of 1 amplifier
- V_S = Total supply voltage
- I_{qMAX} = Maximum quiescent supply current of 1 amplifier
- V_{OUTMAX} = Maximum output voltage swing of the application
- R_L = Load resistance

ISL28108, ISL28208, ISL28408 SPICE Model

Figure 67 shows the SPICE model schematic and Figure 68 shows the net list for the SPICE model. The model is a simplified version of the actual device and simulates important AC and DC parameters. AC parameters incorporated into the model are: $1/f$ and flat band noise voltage, Slew Rate, CMRR, Gain and Phase. The DC parameters are I_{OS} , total supply current and output voltage swing. The model uses typical parameters given in the "Electrical Specifications" Table beginning on page 6. The AVOL is adjusted for 122dB with the dominant pole at 1Hz. The CMRR is set 128dB, $f = 6$ kHz. The input stage models the actual device to present an accurate AC representation. The model is configured for ambient temperature of +25°C.

Figures 69 through 83 show the characterization vs simulation results for the Noise Voltage, Open Loop Gain Phase, Closed Loop Gain vs Frequency, Gain vs Frequency vs R_L , CMRR, Large Signal 10V Step Response, Small Signal 0.05V Step and Output Voltage Swing ± 15 V supplies.

LICENSE STATEMENT

The information in this SPICE model is protected under the United States copyright laws. Intersil Corporation hereby grants users of this macro-model hereto referred to as "Licensee", a nonexclusive, nontransferable licence to use this model as long as the Licensee abides by the terms of this agreement. Before using this macro-model, the Licensee should read this license. If the Licensee does not accept these terms, permission to use the model is not granted.

The Licensee may not sell, loan, rent, or license the macro-model, in whole, in part, or in modified form, to anyone outside the Licensee's company. The Licensee may modify the macro-model to suit his/her specific applications, and the Licensee may make copies of this macro-model for use within their company only.

This macro-model is provided "AS IS, WHERE IS, AND WITH NO WARRANTY OF ANY KIND EITHER EXPRESSED OR IMPLIED, INCLUDING, BUT NOT LIMITED TO, ANY IMPLIED WARRANTIES OF MERCHANTABILITY AND FITNESS FOR A PARTICULAR PURPOSE."

In no event will Intersil be liable for special, collateral, incidental, or consequential damages in connection with or arising out of the use of this macro-model. Intersil reserves the right to make changes to the product and the macro-model without prior notice.

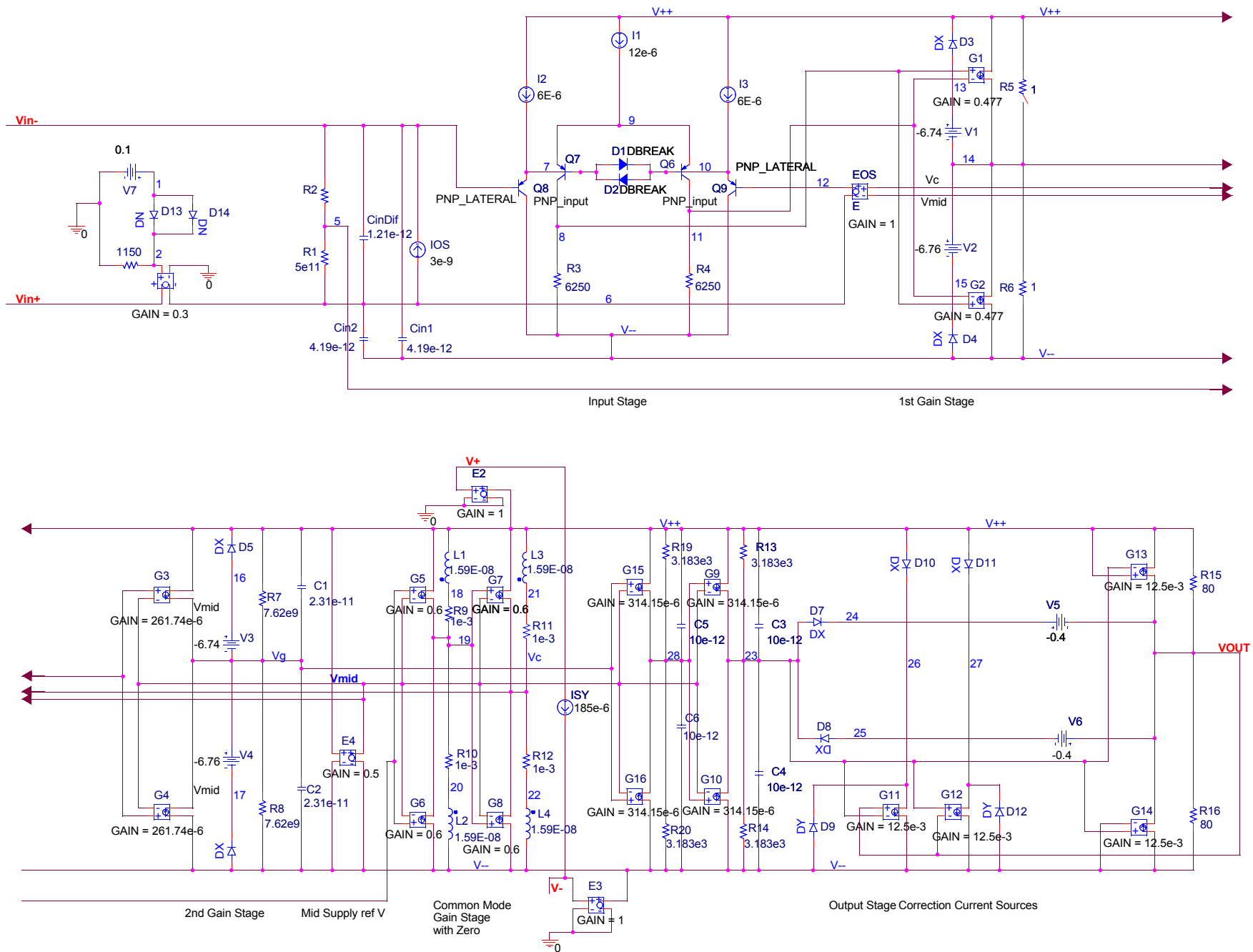


FIGURE 67. SPICE MODEL SCHEMATIC

*ISL28108_208 Macromodel - covers following
*products

*ISL28108
*ISL28208
*ISL28408

*
*Revision History:
* Revision B, LaFontaine January 22 2014
* Model for Noise, supply currents, CMRR
*128dB f=6kHz, AVOL 122dB f=1Hz
* SR = 0.45V/us, GBWP 1.2MHz.
*Copyright 2011 by Intersil Corporation
*Refer to data sheet "LICENSE STATEMENT"
*Use of this model indicates your acceptance
*with the terms and provisions in the License
*Statement.
*

*Intended use:

*This Pspice Macromodel is intended to give
*typical DC and AC performance characteristics
*under a wide range of external circuit
*configurations using compatible simulation
*platforms – such as iSim PE.
*

*Device performance features supported by this
*model

*Typical, room temp., nominal power supply
*voltages used to produce the following
*characteristics:

*Open and closed loop I/O impedances,
*Open loop gain and phase,
*Closed loop bandwidth and frequency
*response,

*Loading effects on closed loop frequency
*response,

*Input noise terms including 1/f effects,

*Slew rate,

*Input and Output Headroom limits to I/O

*voltage swing,

*Supply current at nominal specified supply
*voltages.

*
*Device performance features NOT supported
*by this model:

*Harmonic distortion effects,

*Output current limiting (current will limit at
*40mA),

*Disable operation (if any),

*Thermal effects and/or over temperature

*parameter variation,

*Limited performance variation vs. supply

*voltage is modeled,

*Part to part performance variation due to

*normal process parameter spread,

*Any performance difference arising from

*different packaging source,

*Load current reflected into the power supply
*current.

*
* Connections: +input
* | -input
* | | +Vsupply
* | | -Vsupply
* | | | output
* | | | |
* | | | |
* .subckt ISL28108_208 Vin+ Vin-V+ V- VOUT
* source ISL28118_218_subckt_check_0
*

*Voltage Noise

E_En VIN+ 6 2 0 0.3

D_D13 1 2 DN

D_D14 1 2 DN

V_V7 1 0 0.1

R_R17 2 0 1150

*

*Input Stage

Q_Q6 11 10 9 PNP_input

Q_Q7 8 7 9 PNP_input

Q_Q8 V-- VIN- 7 PNP_LATERAL

Q_Q9 V-- 12 10 PNP_LATERAL

I_I1 V++ 9 DC 12e-6

I_I2 V++ 7 DC 6E-6

I_I3 V++ 10 DC 6E-6

I_IOS 6 VIN- DC 3e-9

D_D1 7 10 DBREAK

D_D2 10 7 DBREAK

R_R1 5 6 5e11

R_R2 VIN- 5 5e11

R_R3 V-- 8 6250

R_R4 V-- 11 6250

C_Cin1 V-- VIN- 4.19e-12

C_Cin2 V-- 6 4.19e-12

C_CinDif 6 VIN- 1.21E-12

*

*1st Gain Stage

G_G1 V++ 14 8 11 0.4779867

G_G2 V-- 14 8 11 0.4779867

V_V1 13 14 -6.74

V_V2 14 15 -6.76

D_D3 13 V++ DX

D_D4 V-- 15 DX

R_R5 14 V++ 1

R_R6 V-- 14 1

*

*2nd Gain Stage

G_G3 V++ VG 14 VMID 261.748e-6

G_G4 V-- VG 14 VMID 261.748e-6

V_V3 16 VG -6.74

V_V4 VG 17 -6.76

D_D5 16 V++ DX

D_D6 V-- 17 DX

R_R7 VG V++ 7.62283e9

R_R8 V-- VG 7.62283e9

C_C1 VG V++ 2.31e-11

C_C2 V-- VG 2.31e-11

*

*Mid supply Ref

E_E2 V++ 0 V+ 0 1

E_E3 V-- 0 V- 0 1

E_E4 VMID V-- V++ V-- 0.5

I_ISY V+ V- DC 185E-6

*

*Common Mode Gain Stage with Zero

G_G5 V++ 19 5 VMID 0.6

G_G6 V-- 19 5 VMID 0.6

G_G7 V++ VC 19 VMID 0.6

G_G8 V-- VC 19 VMID 0.6

E_EOS 12 6 VC VMID 1

L_L1 18 V++ 1.59159E-08

L_L2 20 V-- 1.59159E-08

L_L3 21 V++ 1.59159E-08

L_L4 22 V-- 1.59159E-08

R_R9 19 18 1e-3

R_R10 20 19 1e-3

R_R11 VC 21 1e-3

R_R12 22 VC 1e-3

*

*Pole Stage

G_G15 V++ 28 VG VMID 314.15e-6

G_G16 V-- 28 VG VMID 314.15e-6

R_R19 28 V++ 3.18319e3

R_R20 V-- 28 3.18319e3

C_C5 28 V++ 10e-12

C_C6 V-- 28 10e-12

*

G_G9 V++ 23 28 VMID 314.15e-6

G_G10 V-- 23 28 VMID 314.15e-6

R_R13 23 V++ 3.18319e3

R_R14 V-- 23 3.18319e3

C_C3 23 V++ 10e-12

C_C4 V-- 23 10e-12

*

*Output Stage with Correction Current Sources

G_G11 26 V-- VOUT 23 12.5e-3

G_G12 27 V-- 23 VOUT 12.5e-3

G_G13 VOUT V++ V++ 23 12.5e-3

G_G14 V-- VOUT 23 V-- 12.5e-3

D_D7 23 24 DX

D_D8 25 23 DX

D_D9 V-- 26 DY

D_D10 V++ 26 DX

D_D11 V++ 27 DX

D_D12 V-- 27 DY

V_V5 24 VOUT -0.4

V_V6 VOUT 25 -0.4

R_R15 VOUT V++ 80

R_R16 V-- VOUT 80

.model PNP_LATERAL pnp(is=1e-016 bf=250
va=80

+ ik=0.138 rb=0.01 re=0.101 rc=180 kf=0 af=1)

.model PNP_input pnp(is=1e-016 bf=100

va=80

+ ik=0.138 rb=0.01 re=0.101 rc=180 kf=0 af=1)

.model DBREAK D(bv=43 rs=1)

.model DN D(KF=6.69e-9 AF=1)

.MODEL DX D(IS=1E-12 Rs=0.1)

.MODEL DY D(IS=1E-15 BV=50 Rs=1)

.ends ISL28108_208

FIGURE 68. SPICE NET LIST

Characterization vs Simulation Results

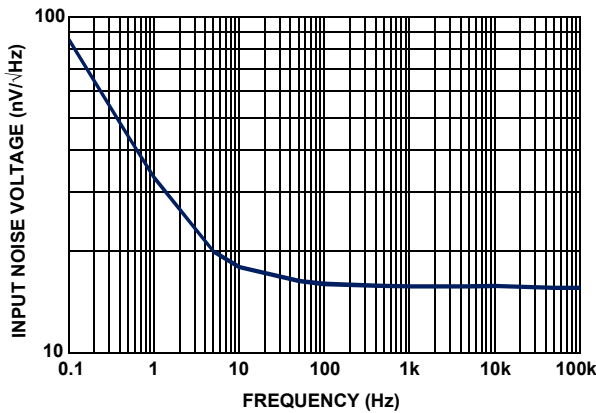


FIGURE 69. CHARACTERIZED INPUT NOISE VOLTAGE

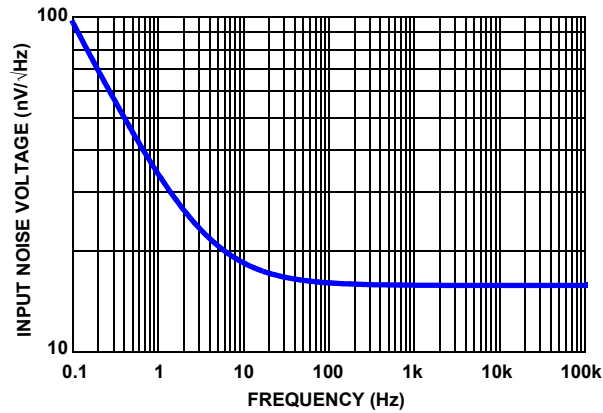


FIGURE 70. SIMULATED INPUT NOISE VOLTAGE

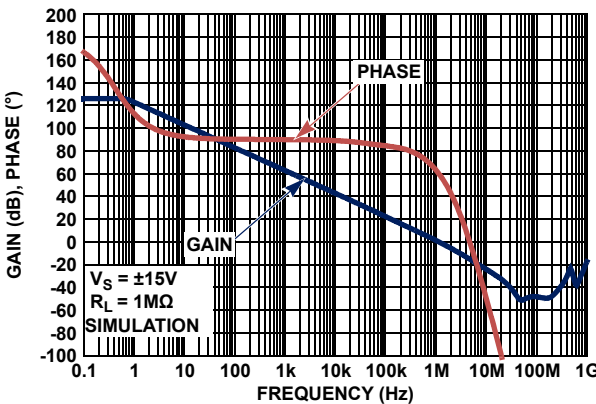


FIGURE 71. CHARACTERIZED OPEN-LOOP GAIN, PHASE vs FREQUENCY

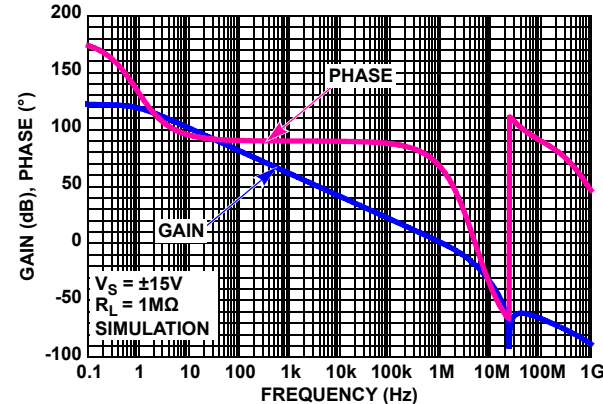


FIGURE 72. SIMULATED OPEN-LOOP GAIN, PHASE vs FREQUENCY

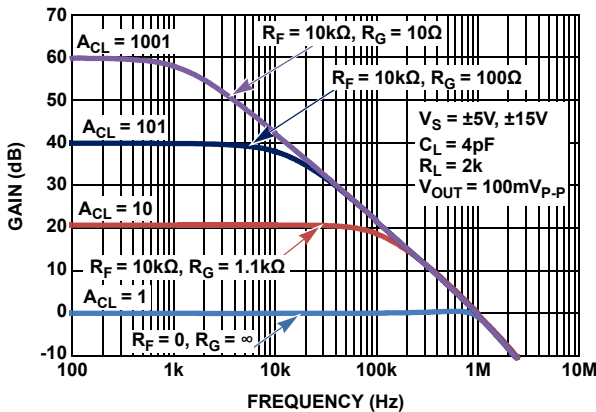


FIGURE 73. CHARACTERIZED CLOSED LOOP GAIN vs FREQUENCY

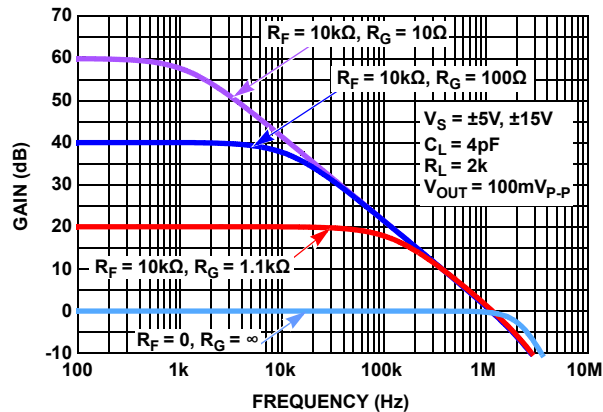


FIGURE 74. SIMULATED CLOSED LOOP GAIN vs FREQUENCY

Characterization vs Simulation Results (Continued)

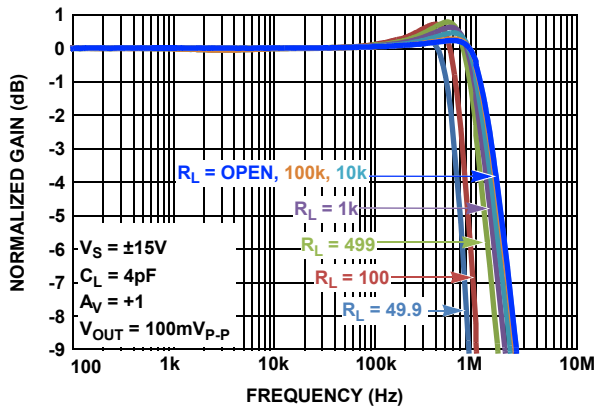


FIGURE 75. CHARACTERIZED GAIN vs FREQUENCY vs R_L

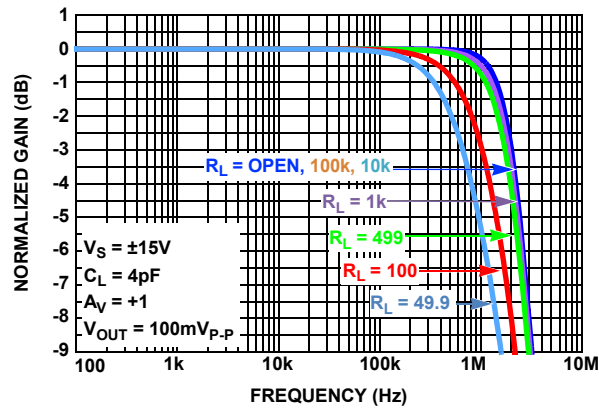


FIGURE 76. SIMULATED GAIN vs FREQUENCY vs R_L

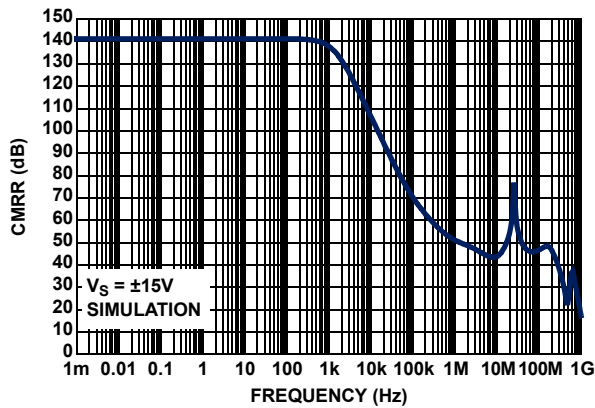


FIGURE 77. CHARACTERIZED CMRR vs FREQUENCY

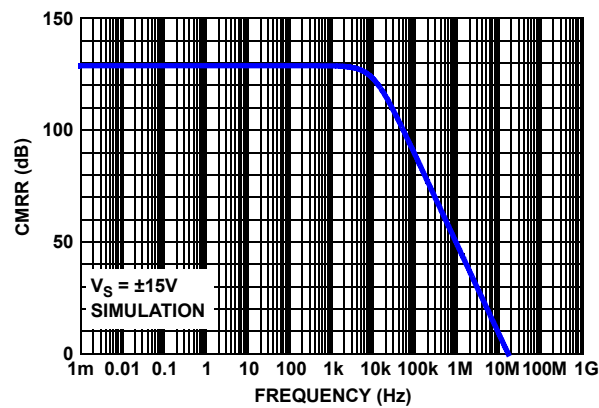


FIGURE 78. SIMULATED CMRR vs FREQUENCY

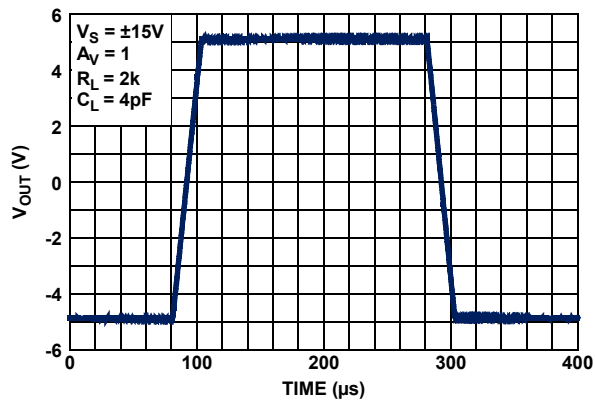


FIGURE 79. CHARACTERIZED LARGE SIGNAL 10V STEP RESPONSE

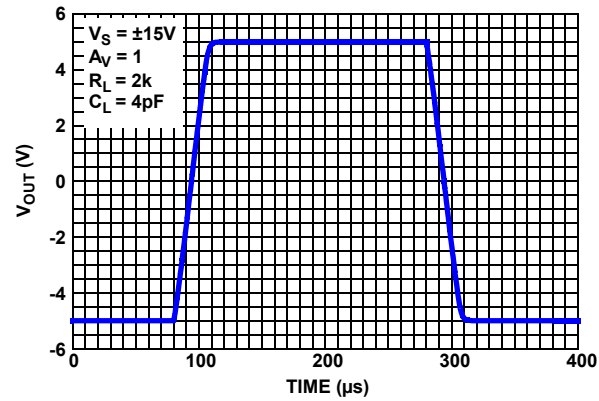


FIGURE 80. SIMULATED LARGE SIGNAL 10V STEP RESPONSE

Characterization vs Simulation Results (Continued)

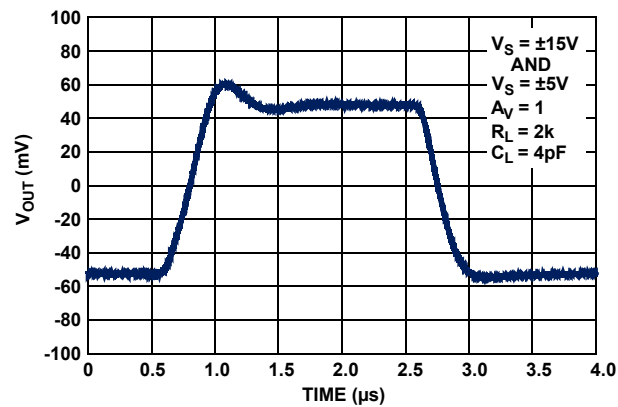


FIGURE 81. CHARACTERIZED SMALL SIGNAL TRANSIENT RESPONSE

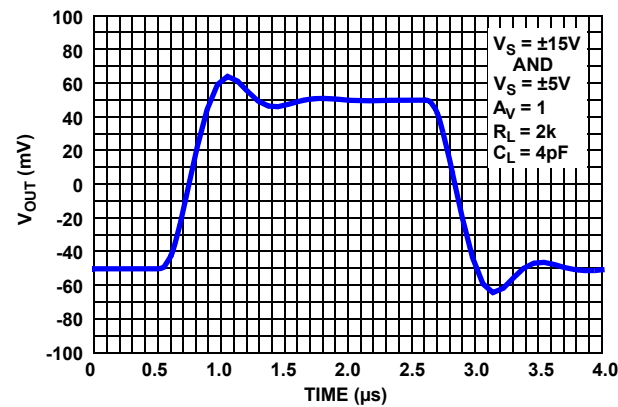


FIGURE 82. SIMULATED SMALL SIGNAL TRANSIENT RESPONSE

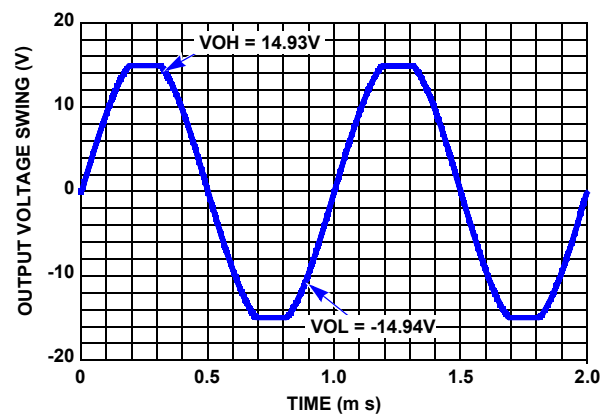


FIGURE 83. SIMULATED OUTPUT VOLTAGE SWING

Revision History

The revision history provided is for informational purposes only and is believed to be accurate, but not warranted. Please go to web to make sure you have the latest Rev.

DATE	REVISION	CHANGE
March 4, 2014	FN6935.5	Updated absolute Maximum Ratings table on page 6, by adding ISL28108 MSOP value to ESD tolerance. Updated Spice model netlist on page 24.
January 22, 2013	FN6935.4	Added ISL28208 MSOP specifications. Moved Table of Contents to page 2. Removed "Coming Soon" from ISL28208FUZ entry in "Ordering Information" table on page 3. page 6, removed package indicators on Vos entries for ISL28208 and ISL28408; added ISL28208 MSOP entries for TCvos and deltaVos. page 7, removed package indicators on los entries for ISL28208 and ISL28408. page 8, removed package indicators on Vos and los entries for ISL28208 and ISL28408; added ISL28208 MSOP entries for TCvos, and added "ISL28108, ISL28208" to 4th and 5th rows of CMRR parameter. page 8, added ISL28208 MSOP entries for PSRR. page 10, added "+25°C" to "default" info at top of page. page 12, added "MSOP" to titles of Figures 15 and 16.
October 19, 2011	FN6935.3	- On page 6, ±15V Electrical Specifications table, added the following parameters: - VOS ISL28108 SOIC, TDFN - TCvos ISL28108 SOIC, TDFN - IOS ISL28108 SOIC, TDFN - On page 6 and 6, ±15V Electrical Specifications table, changed the following in Conditions column: - VOS: changed "ISL28208 SOIC, TDFN" to "ISL28208 SOIC, TDFN; ISL28408 SOIC" - TCvos: changed "ISL28208 TDFN" to "ISL28208 TDFN, ISL28408 SOIC" - IOS: changed "ISL28108 SOIC, TDFN" to "ISL28108 SOIC, TDFN; ISL28408 SOIC" - On page 8, ±5V Electrical Specifications table, added the following: - VOS ISL28108 SOIC, TDFN - TCvos ISL28108 SOIC, TDFN - IOS ISL28108 SOIC, TDFN - On page 8, ±5V Electrical Specifications table, changed the following in Conditions column: - VOS: changed "ISL28208 SOIC, TDFN" to "ISL28208 SOIC, TDFN; ISL28408 SOIC" - TCvos: changed "ISL28208 TDFN" to "ISL28208 TDFN, ISL28408 SOIC" - IOS: changed "ISL28108 SOIC, TDFN" to "ISL28108 SOIC, TDFN; ISL28408 SOIC" - On page 10 through page 12, added the following to Typical Performance Curves: - Figures 3, 4: ISL28408 SOIC Input Offset Distribution Voltage, ±15V and ±5V - Figures 7, 8: ISL28108 SOIC ±15V VOS distribution, and ±5V VOS distribution - Figures 9, 10: ISL28108 TDFN ±15V VOS distribution, and ±5V VOS distribution - Figures 11, 12: ISL28408 SOIC TCvos vs. number of Amplifiers ±15V and ±5V - Figures 17, 18: ISL28108 SOIC ±15V TCvos distribution, and ±5V TCvos distribution - Figures 19, 20: ISL28108 TDFN ±15V TCvos distribution, and ±5V TCvos distribution - On page 21, Applications: minor edits to re-align figures with curves - On page 22: changed heading from "Using Only One Amplifier to "Unused Channels" and edited for clarity. - On page 32: changed Package Outline Drawing L8.3x3A to L8.3x3K - On page 33: changed Package Outline Drawing M8.118 to M8.118B - On page 34: added Package Outline Drawing M14.15 - On page 1, Features: changed Low Input Offset Voltage from 230µV to (ISL28108).....150µV. Added Related Literature section with AN1658, "ISL28208SOICEVAL2Z Evaluation Board User Guide". - On page 3, Ordering Information: added ISL28208SOICEVAL2Z evaluation board. Removed "Coming Soon" from ISL28408FBZ. Corrected Package Dwg. # for TDFN package from L8.3x3A to L8.3x3K. Corrected Package Dwg. # for MSOP package from M8.118 to M8.118B. - On page 6, Absolute Maximum Ratings, changed "ESD Tolerance (ISL28208)" to "ESD Tolerance (ISL28208, ISL28408)". Added ESD information for ISL28108 as follows: ESD Tolerance (ISL28108) Human Body Model (Tested per JESD22-A114F).....5.5kV Machine Model (Tested per JESD22-A115-C).....300V Charged Device Model (Tested per JESD22-C110D).....2kV

Revision History

The revision history provided is for informational purposes only and is believed to be accurate, but not warranted. Please go to web to make sure you have the latest Rev.

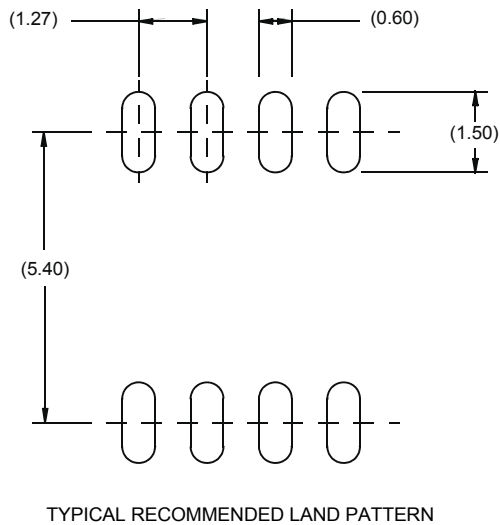
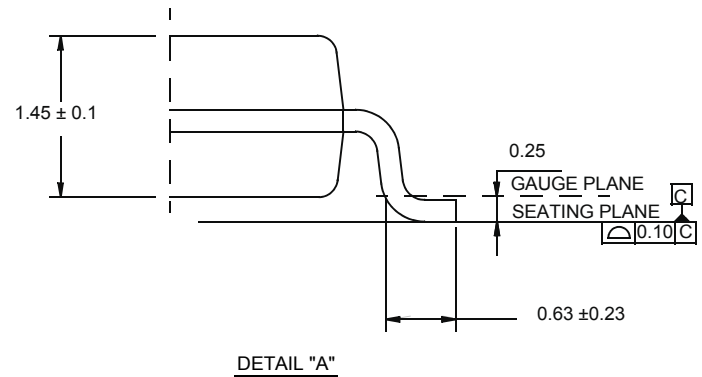
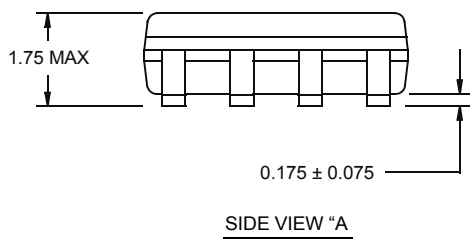
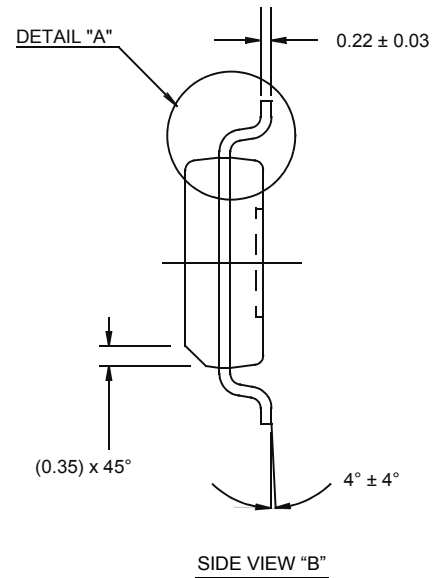
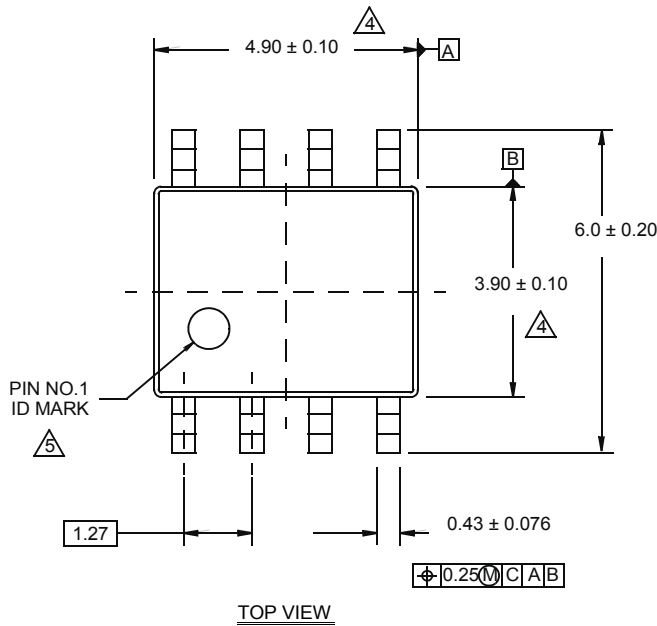
DATE	REVISION	CHANGE
October 19, 2011	FN6935.3	On page 6, Thermal Information, changed package temperatures from: 8 Ld SOIC Package (108, 208, Notes 4, 7), $\theta_{JA} = 120$, $\theta_{JC} = 55$ 8 Ld TDFN Package (108, 208, Notes 5, 6), $\theta_{JA} = 47$, $\theta_{JC} = 6$ 8 Ld MSOP Package (108, 208, Notes 4, 7), $\theta_{JA} = 150$, $\theta_{JC} = 45$ 14 Ld SOIC Package (408, Notes 4, 7), $\theta_{JA} = -$, $\theta_{JC} = -$ To: 8 Ld SOIC Package (208, Notes 4, 7), $\theta_{JA} = 120$, $\theta_{JC} = 55$ and 8 Ld SOIC Package (108, Notes 4, 7), $\theta_{JA} = 120$, $\theta_{JC} = 60$ 8 Ld TDFN Package (208, Notes 5, 6), $\theta_{JA} = 47$, $\theta_{JC} = 6$ and 8 Ld TDFN Package (108, Notes 5, 6), $\theta_{JA} = 45$, $\theta_{JC} = 3.5$ 8 Ld MSOP Package (208, Notes 4, 7), $\theta_{JA} = 150$, $\theta_{JC} = 50$ and 8 Ld MSOP Package (108, Notes 4, 7), $\theta_{JA} = 165$, $\theta_{JC} = 57$ 14 Ld SOIC Package (408, Notes 4, 7), $\theta_{JA} = 71$, $\theta_{JC} = 37$
April 20, 2011	FN6935.2	- Added discussion of ISL28408 throughout datasheet. - On page 3 in "Ordering Information": Added new part, "ISL28408FBZ". Corrected part marking for ISL28208FRTZ from 208Z to 208F. Added "ISL28408" to Note 3. Under "Pin Configurations," added ISL28408 (14 Ld SOIC) pin configuration diagram. - On page 5: in Pin Descriptions table, added column for ISL28408 14Ld SOIC. Corrected schematic for Circuit 2. - On page 6: under "Thermal Information" added "14 Ld SOIC Package (408, Notes 4, 7)" and added ISL28108 to 8 Ld TDFN and 8 Ld MSOP. Changed θ_{JA} and θ_{JC} for 8 Ld TDFN Package from 48 and 5.5 to 47 and 6. Added Note 6 regarding θ_{JC} "case temp" measurement, and applied it to 8 Ld TDFN Package. - On page 6: in Electrical Specifications table, changed TYP spec for TCI_B from 70 pA/° C to 0.07nA/° C. On page 8, changed TYP spec for TCI_B from -67 pA/° C to -0.067nA/° C. These are not spec changes since the values are the same. - On page 14, Figs. 31 and 32: changed y axis units label from (mV) to (V); changed x axis units label from (μA) to (mA). - On page 21, under "Output Drive Capability," para 2, changed "The output stage can swing at moderate levels of output current (Figures 21 and 22) and the output stage is internally current limited. Output current limit over-temperature..." to "The output stage is internally current limited. Output current limit over-temperature..."
March 11, 2011	FN6935.1	- On page 1, in the first paragraph - added the following after V-rail: "a rail-to-rail differential input voltage range for use as a comparator,..." - On page 1 in "Features": - Added bullet - "Rail-to-rail input differential voltage range for comparator applications" - Changed Low Noise Current from "100fA/sq.root Hz" to "80fA/sq.root Hz" - On page 3 in "Ordering Information" - Removed "coming soon" from ISL28208FRTZ part since it is releasing. - On page 6, changed "ESD Tolerance (ISL28208, ISL28408)" as follows: - Human Body Model changed from "3kV" to "6kV" - Machine Model changed from "300V" to "400V" - Added JEDEC Test information for all ESD ratings - On page 6 and page 8, added test conditions for SOIC TCVoS specs. Added TCVoS specs for TDFN. - On page 7 changed "Noise Current Density" Typical from "100" to "80" - On page 21, updated Applications Information Functional Description - On page 21 Updated Input Stage Performance Section - On page 21 Updated Output Drive Capability Section - On page 22 Added ISL28108 AND ISL28208 SPICE MODEL and License Agreement section - On page 23 Added SPICE NET LIST - On page 25 Added Characterization vs Simulation Results curves
February 16, 2011	FN6935.0	Initial Release

Package Outline Drawing

M8.15E

8 LEAD NARROW BODY SMALL OUTLINE PLASTIC PACKAGE

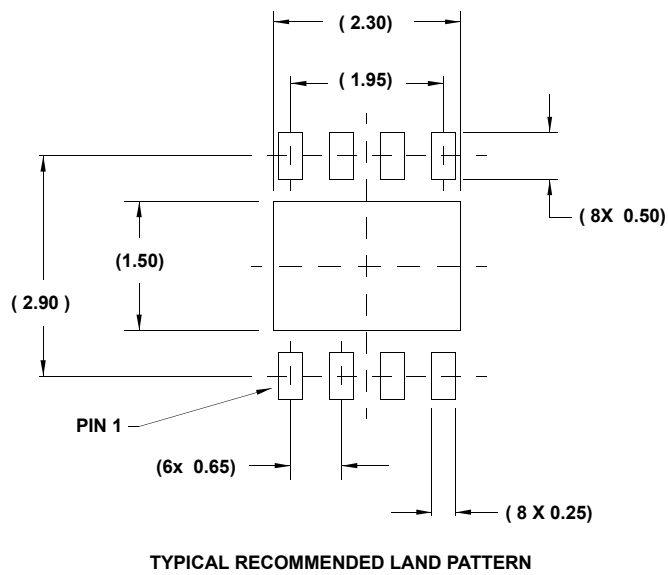
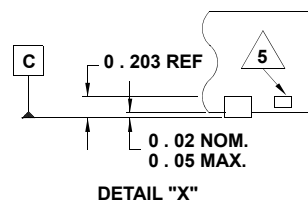
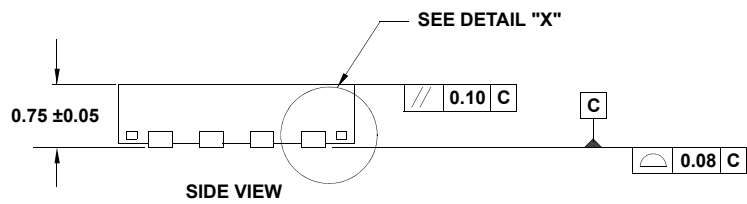
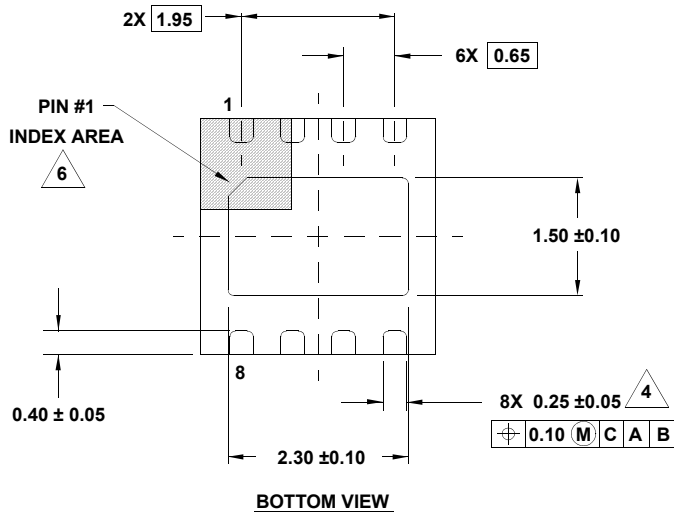
Rev 0, 08/09



NOTES:

1. Dimensions are in millimeters.
Dimensions in () for Reference Only.
2. Dimensioning and tolerancing conform to AMSE Y14.5m-1994.
3. Unless otherwise specified, tolerance : Decimal ± 0.05
4. Dimension does not include interlead flash or protrusions.
Interlead flash or protrusions shall not exceed 0.25mm per side.
5. The pin #1 identifier may be either a mold or mark feature.
6. Reference to JEDEC MS-012.

Rev 1, 9/11



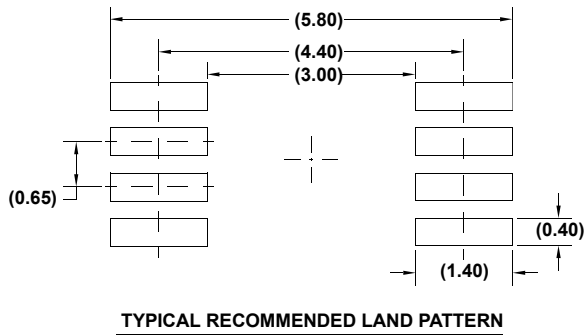
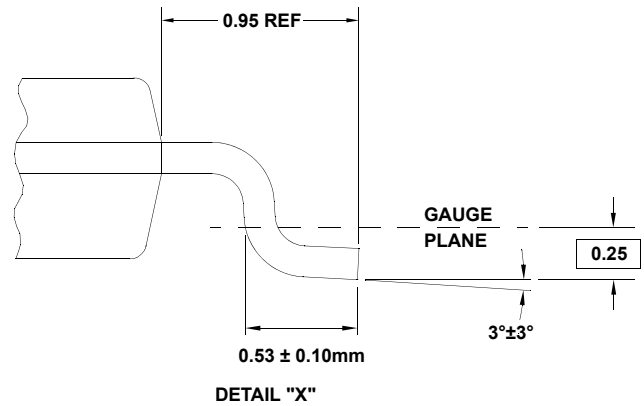
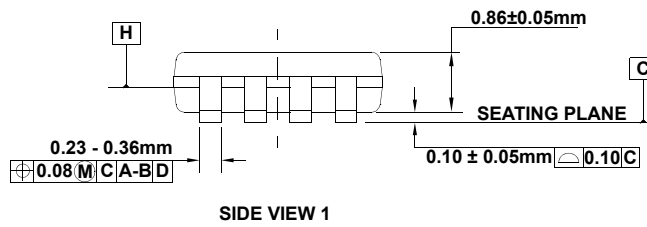
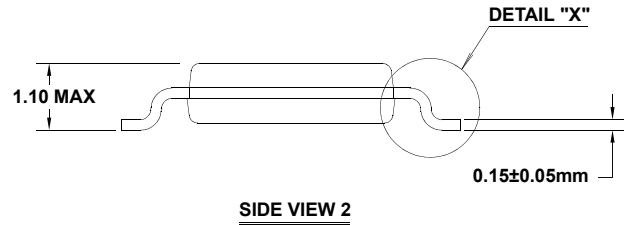
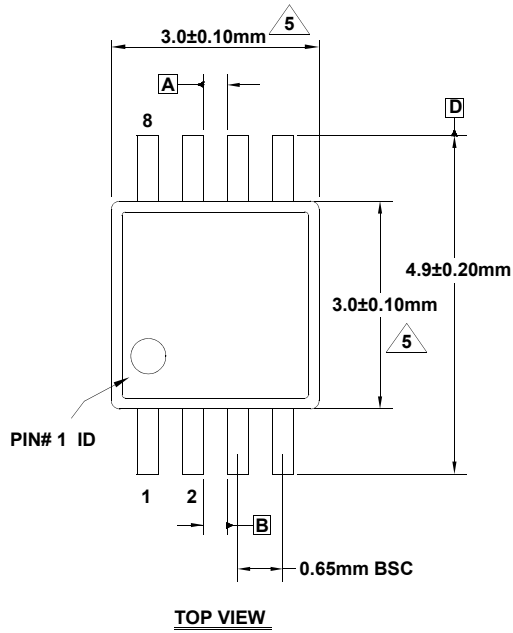
1. Dimensions are in millimeters.
Dimensions in () for Reference Only.
2. Dimensioning and tolerancing conform to AMSE Y14.5m-1994.
3. Unless otherwise specified, tolerance : Decimal ± 0.05
4. Dimension applies to the metallized terminal and is measured between 0.15mm and 0.20mm from the terminal tip.
5. Tiebar shown (if present) is a non-functional feature.
6. The configuration of the pin #1 identifier is optional, but must be located within the zone indicated. The pin #1 identifier may be either a mold or mark feature.
7. Compliant to JEDEC MO-229 WEEC-2 except for the foot length.

Package Outline Drawing

M8.118B

8 LEAD MINI SMALL OUTLINE PLASTIC PACKAGE

Rev 1, 3/12

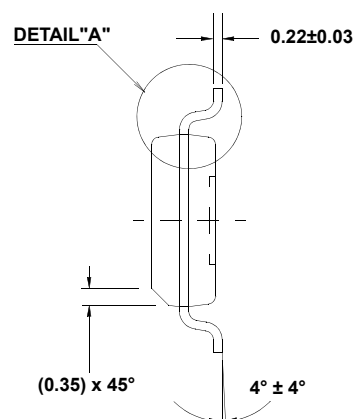


NOTES:

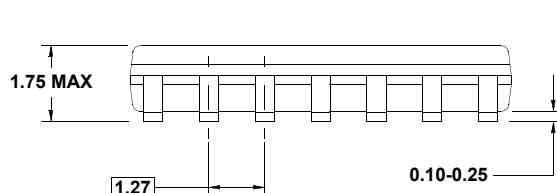
1. Dimensions are in millimeters.
2. Dimensioning and tolerancing conform to JEDEC MO-187-AA and AMSEY14.5m-1994.
3. Plastic or metal protrusions of 0.15mm max per side are not included.
4. Plastic interlead protrusions of 0.15mm max per side are not included.
5. Dimensions are measured at Datum Plane "H".
6. Dimensions in () are for reference only.

M14.15

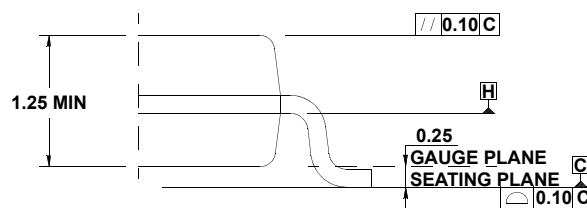
Rev 1, 10/09



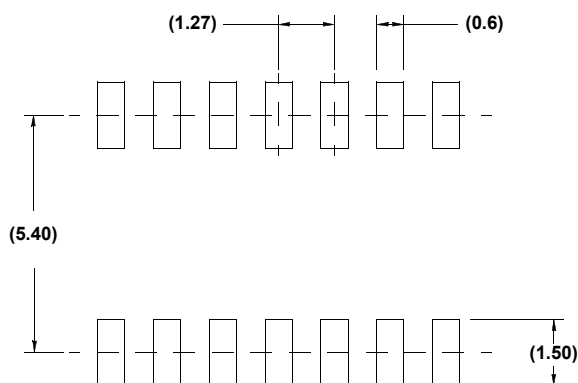
TOP VIEW



SIDE VIEW



DETAIL "A"



TYPICAL RECOMMENDED LAND PATTERN

NOTES:

1. Dimensions are in millimeters.
Dimensions in () for Reference Only.
2. Dimensioning and tolerancing conform to AMSEY14.5m-1994.
3. Datums A and B to be determined at Datum H.
4. Dimension does not include interlead flash or protrusions.
Interlead flash or protrusions shall not exceed 0.25mm per side.
5. The pin #1 identifier may be either a mold or mark feature.
6. Does not include dambar protrusion. Allowable dambar protrusion shall be 0.10mm total in excess of lead width at maximum condition.
7. Reference to JEDEC MS-012-AB.